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"HIS MASTER'S VOICE"

SERVICE MANUAL

for

TELEVISION RECEIVER

CHASSIS TYPES M3, M4, M5, M6, M7, P1, P2, P5, P6, P7 and P8

THE GRAMOPHONE COMPANY LIMITED (Inc. in England) HOMEBUSH - N.S.W.

PART No. 683-0831

INTRODUCTION

This combined service manual is intended to give the serviceman, within the one cover, a complete coverage of all of the present series of "H.M.V" 110-degree receivers and provide him with all of the necessary information for servicing these receivers. It is anticipated that by combining this information, that would normally be written in separate manuals, that a more complete understanding of the variations in circuitry between these receivers will result, and that servicing will thus be simplified.

It will be seen that the basic chassis may be divided into 13 or 15 valve receivers; that the method of mounting is divided into cabinet or transportable receivers; that a pre-set tone control is provided only on 15-valve 23-inch receivers; and that the remote control facility is confined to 15-valve receivers.

Chassis Type	No. of Valves	Picture Tube	Style of Receiver	Remote Control	Tone Control
M3	 15	21 inch	Transportable	Yes	
M4	 13	21 inch	Wooden Cabinet		
M5	 13	21 inch	Transportable		
M6	 15	21 inch	Wooden Cabinet	Yes	
M7	 13	21 inch	Wooden Consolette		
P1	 15	23 inch	Wooden Cabinet	Yes	Yes
P2	 15	23 inch	Transportable	Yes	Yes
P5	 15	23 inch	Deep Cabinet	Yes	Yes
P6	 13	23 inch	Transportable Wooden Cabinet		
P7	 13	23 inch	Deep Cabinet. Transportable or Wooden Consolette		
P8	 13	23 inch	Transportable		

The receivers covered in this manual are:

CAUTION

The normal B+ voltages in these receivers are dangerous. Use extreme caution when servicing. The high voltage at the picture tube anode (16,000 volts) will give an unpleasant shock but does not supply enough current to give a fatal burn or shock. However, secondary human reactions to otherwise harmless shocks have been known to cause injury.

Always discharge the picture tube anode to the chassis or to its aquadag coating before handling the tube. The picture tube is highly evacuated and if broken it will violently expel glass fragments. When handling the picture tube, always wear goggles.

SPECIFICATIONS

POWER SUPPLY:

200, 230, 240, 250 volts, A.C., 50 c.p.s.

CONSUMPTION:

15 valve receivers — 180 watts. 13 valve receivers — 170 watts.

AERIAL INPUT:

300 ohms balanced.

Transportable receivers have additional inbuilt aerial and provision for a plug-in attenuator.

INTERMEDIATE FREQUENCIES:

Vision carrier — 36.875 Mc/s. Sound carrier — 31.375 Mc/s.

FUSES:

Mains — 1 amp. H.T.1 — 1.5 amp. H.T.2 — 250 mA.

VALVE COMPLEMENTS

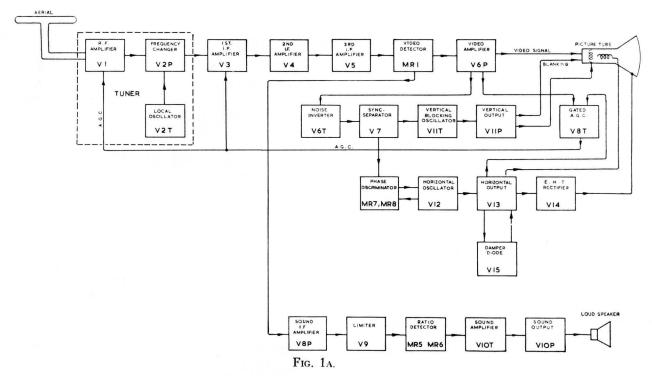
15 Valve Receivers-M3, M6, P1, P2 and P5.

V1 V2 V3 V4 V5 V6 V7 V8	6ES8 6BL8 6EH7 6EJ7 6EJ7 6DX8 6DT6 6U8	 R.F. Amplifier Frequency Changer 1st I.F. Amplifier 2nd I.F. Amplifier 3rd I.F. Amplifier Video Amplifier and Noise Inverter Noise Gated Sync. Separator Sound I.F. Amplifier and 	V12 V13 V14 V15 MR1 MR2 MR3 MR4 MR5	12AU7 6CM5 1S2 6AL3 0A80 0A210 0A210 M3 0A79	Horizontal Multivibrator Horizontal Output EHT Rectifier Damping Diode Vision Detector Mains Rectifier Mains Rectifier Clamping Diode Ratio Detector
V8 V9 V10 V11	6U8 6AU6 6BM8 6BM8	Sound I.F. Amplifier and Gated A.G.C. Sound Limiter Audio Driver and Audio Output Blocking Oscillator and Verti- cal Output	MR5 MR6 MR7 MR8	0A79 0A79 0A81 0A81	Ratio Detector Ratio Detector Phase Discriminator Phase Discriminator

13 Valve Receivers-M4, M5, M7, P6, P7 and P8

V1	6ES8	R.F. Amplifier	V10	12AU7	Horizontal Multivibrator
V2	6BL8	Frequency Changer	V11	6CM5	Horizontal Output
V3	6BY7	1st I.F. Amplifier	V12	1S2	EHT Rectifier
V4	6BX6	2nd I.F. Amplifier	V13	6AL3	Damping Diode
V5	6U8	3rd I.F. Amplifier and Noise	MR1	0A80	Vision Detector
		Inverter	MR2	OA210	Mains Rectifier
V6	6DX8	Video Amplifier and Sync.	MR3	0A210	Mains Rectifier
	CLIO	Separator	MR4	0A81	Pre-clamping Diode
V7	6U8	Sound Limiter and Gated A.G.C.	MR5	M3	A.G.C. Clamping Diode
V8	6BM8	Audio Driver and Audio	MR6	0A79	Ratio Detector
vo	ODMO	Output	MR7	0A79	Ratio Detector
V9	6BM8	Blocking Oscillator and Verti-	MR8	0A81	Phase Discriminator
		cal Output	MR9	0A81	Phase Discriminator

The Cathode ray tubes employed in these receivers are shown under the heading, "Picture Tube Replacement," on Page 16.



BLOCK DIAGRAM — 15-VALVE CIRCUITS.

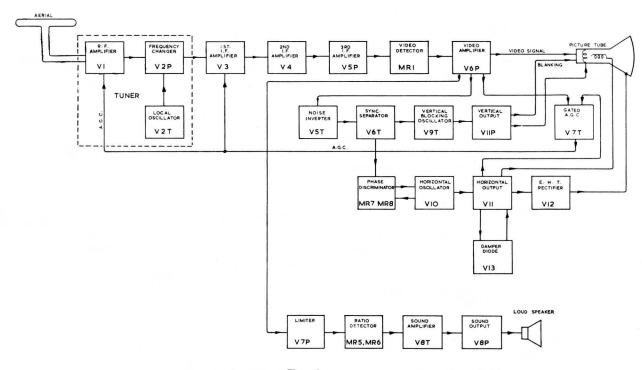


Fig. 1b. BLOCK DIAGRAM — 13-VALVE CIRCUITS.

SUMMARY OF FEATURES

These features are common to both types of receiver. In later paragraphs features that differ in each type of receiver are listed, so that the main differences are pointed out.

1. The turret tuner has facilities for individual exact alignment, on each channel, of the oscillator tuning so that the fine tuning control range may be limited, thus avoiding gross mistuning.

2: Linear phase treatment of the IF response ensures the best possible definition with freedom from overshoot or smearing, allowing non-critical fine tuning control.

3. The overall frequency response of the system is within 6 db from D.C. up to 4.7 Mc/s.

4. DC coupling from the video detector through to the picture tube ensures that a true black is retained and that all shades retain their true relationship to black. This prevents fading to grey and gives accurate portrayal of night-time scenes.

5. Time-gated A.G.C. is employed, giving immunity from the effects of impulse noise and has a fast action to cope with rapid fading from "aircraft flutter." A variable delay on the tuner is provided to maintain full RF gain on weak to moderate signals, thus minimising frequency converter noise. This delay can be adjusted for best results when the receiver is first installed.

6. A noise inverter is used, before the sync. separator, giving protection to the input circuit of the sync. separator in the presence of impulse noise, preventing paralysis of the sync. separator action following large bursts of impulse noise.

7. The audio amplifier with ample feedback gives excellent quality sound with ample power. A compensated volume control maintains tonal balance at all volume settings.

8. The horizontal hold circuit is a multivibrator employing a stabilising coil in the oscillator anode circuit and this, together with the cathode coupling employed, gives a stable oscillator virtually unaffected by large HT variations. This stability, together with an adequate pull-in range, renders a front horizontal hold control unnecessary. A pre-set control is provided on the back of the receiver.

9. A linearity control of the horizontal deflection circuit that can be adjusted by an indication on a multimeter gives the advantage that the linearity can be set without the need for a transmitted test pattern. 10. Vertical retrace lines are eliminated by Vertical Flyback Blanking.

11. Dustproof seal around picture tube to eliminate dust which would, otherwise, be attracted by static charge.

12. The user controls are reduced to the minimum necessary to ensure correct operation, which is made as simple as possible. There are no interacting controls, and since the receiver is completely DC coupled then the Brightness control will vary the light output of the picture tube and the Contrast control will vary the depth of shades without any interaction between them.

13. Silicon Junction Diodes are used in a voltage doubler full-wave power supply. These diodes are noted for their robustness, economy of power and high surge rating. Since they have no heated cathode they cannot be damaged by operation on low mains voltage as is the case with valve rectifiers. Thermistor protection is included in the power supply to ensure that diodes and electrolytic capacitors are not subjected to large surges when first switching on the receiver.

14. The chassis is hinged and can be swung out in such a manner that almost any repair can be made in the home without removing the whole chassis from the cabinet. It is also possible to remove the whole receiver, complete with the picture tube, from the cabinet in one piece as a complete working unit, and can be operated as such on the workshop bench.

The following features are not common to both models and the differences are well worth noting so that circuitry differences can be checked on the relevant circuit diagrams.

1. The 15-valve receiver uses a noise-gated sync. separator to give the best synchronisation obtainable, necessary for receivers operating under "fringe" conditions. In the 13-valve receivers a triode sync. separator is employed.

2. To keep impulse interference to a minimum in the audio output of the 15-valve receivers, when operating under adverse conditions, a Sound I.F. Amplifier has been included before the Sound Limiter. The substantial increase in overall gain, as compared with the 13-valve model, gives virtually noise - free sound, even under extreme "fringe" conditions. 3. Current feedback, to keep a constant deflection current in the coils, is used in the 15valve receiver Vertical Deflection circuit. A conventional Blumlein circuit is used in the 13-valve receiver but with the added compensation of a thermistor, in series with the Vertical Deflection coils and in direct contact with them. As the coils heat up their resistance increases, but the resistance of the thermistor decreases and compensation is made to keep the deflection current a constant. Both circuits hold the height constant as the deflection coils warm up. Good interlace is maintained in both circuits.

R.F. INPUT

The input to the turret tuner is to a centre tapped transformer which presents an impedance of 300 ohms.

R.F. amplification is achieved with a type 6ES8, double triode (V1), in a cascode circuit. The two sections of this stage are connected in series for DC. The grounded cathode input section is neutralised and is also controllable by A.G.C. from the mains chassis. Because of the series DC connection of the two portions, A.G.C. voltage to one section also effects control on the other section.

Coupling between the two sections of the cascode is direct and the coil between the two maintains amplification on the high frequency channels.

Inductive coupling is used between the cascode and mixer. V2, a type 6BL8, combined triode-pentode, is used as oscillator and mixer. The oscillator is a Colpitts circuit operating above signal frequency. Injection to the mixer input is by inductive coupling. The fine tuning capacitor is capacitively coupled to the oscillator coil by a contact lug on the coil former. Adjustment on each channel is provided by means of a screwed slug in each oscillator coil, this slug being accessible through a hole in the front plate of the tuner when the fine tuning capacitor is in an approximate mid-position.

The fine tuning capacitor takes the form of a specially-shaped ceramic wafer which turns between two fixed metal plates.

The intermediate frequency output of the tuner (vision 36.875 Mc/s., sound 31.375 Mc/s.) is coupled to the IF channel of the main chassis through a secondary winding on the IF coil L7.

The heater circuit is filtered by a Ferroxcube bead through which a heater wire is passed. The bead concentrates the field around the wire, in4. Transformer-coupled focusing is employed on the 15-valve receivers giving good overall edgeto-edge focus. Degradation of definition due to change in spot shape across the tube is thus obviated. Normal DC focusing is employed on the 13-valve models.

5. The 15-valve receiver employs frame-grid valves in the I.F. stages. These valves give more gain than previously used types and ensure that the receiver has ample gain even under the worst signal conditions.

CIRCUIT DESCRIPTION

creasing its self-inductance so that it acts as a choke.

IF AMPLIFIER

The tuner IF output is coupled to the grid of the first IF Amplifier V3 and tuned by coil L9 with stray capacities. There are three IF amplifying stages and AGC voltage is applied to V3.

V5 is coupled to the video detector MR1 by inductive coupling.

Trap circuits L11, L12 and L15 are coupled to the IF coils L10, IFT1 and IFT2. The first attenuates the sound carrier 31.375 Mc/s and the second attenuates the adjacent vision carrier 29.875 Mc/s. The third trap attenuates the adjacent sound carrier 38.375 Mc/s.

V3 and V4 have small unbypassed cathode resistors R15 and R20 to minimise detuning of their grid circuits with varying input levels.

VIDEO AMPLIFIER

The detected video output of the germanium diode MR1 is amplified in the pentode section of V6. L20, L21, L22 and L23 are peaking chokes which maintain the high frequency components of the vision signal fed to the picture tube. The 5.5 Mc/s. component is removed by the combined transformer and trap, IFT4.

INTERCARRIER SOUND

The output of IFT4 is fed to the Sound IF amplifier, in 15-valve receivers, or to the Limiter in 13-valve receivers. The output from the limiter is demodulated by the ratio detector, 2-0A79, to provide the audio signal which passes through the volume control to be amplified in the driver triode and output pentode sections of the Audio Output valve. Feedback is applied in both audio circuits. A full margin of sound gain is provided so that the full 2 watts undistorted output is obtained from sound signals which are not fully modulated. Moreover, the sound output stage has a controlled overload characteristic which does not "paralyse" but merely clips the peaks and so remains comparatively free from audible distortion.

NOISE INVERTER

The anode load of the noise inverter is formed by the resistor coupling the sync. separator to the video output. The valve is biased such that it cannot conduct on the positive sync. tips. However, noise pulses appearing more positive at its grid will drive the valve into conduction, causing current to flow and a voltage drop across this resistor. Consequently a noise pulse will appear less positive, at the anode of the noise inverter, than a sync. tip and the sync. separator will not conduct on the noise pulse since the pulse will now fall outside of the sync. separator's grid base.

NOISE-GATED SYNC. SEPARATOR

(15-value receivers only).

Video signal, with sync. tips positive, is applied to the suppressor grid of a 6DT6 from the output of the video amplifier. At the same time, video signal wth sync. tips negative, is applied to the control grid of this valve, via the potentiometer R45, R46 and is so arranged that the negative excursion of the signal will not affect the current through the sync. separator valve. However, when a noise pulse, which will sit more negative than the sync. tips, occurs at the control grid, then the current through the valve is cut off and the anode voltage will rise to HT, giving no spurious sync. output. Double protection is thus afforded by the Noise Inverter and the Noisegated Sync. Separator.

GATED A.G.C.

Video signals with sync. tips positive are fed from the Video Amplifier anode to the grid of the AGC valve, and the valve is biased so that it will only conduct on sync. tips. During line flyback, a positive pulse is applied to its anode via the 68 pfd coupling capacitor and the valve will conduct when this pulse at its anode and a sync. tip at its grid coincide. The valve cannot, therefore, conduct in the period between sync. pulses and is thus immune to noise pulses appearing in the period between sync. pulses.

The current through the valve will depend on the height of the sync. pulses at its grid and the height of these sync. pulses is adjusted by the contrast control. The cathode of the A.G.C. valve is held at about 50 volts. Operation of the Contrast Control will vary the bias applied to the grid of the A.G.C. valve and so increase or decrease the height of sync. tips in respect to the fixed cathode volts. Increasing the height of the sync. tips will cause the valve to conduct harder and will produce more A.G.C. volts, reducing the gain of the receiver and decreasing the voltage available to drive the C.R.T. cathode. Decreasing the height of the sync. tips will reduce the conduction of the A.G.C. valve, thus producing less A.G.C. volts, increase the gain of the receiver and increase the volts available to drive the C.R.T. cathode. Increasing the height of the sync. tips therefore reduces the contrast, and decreasing the height of the sync. tips will increase the contrast.

The ratio of IF A.G.C. voltage to Tuner A.G.C. voltage is important and the ratio can be adjusted by means of RV3. If the ratio is too small then, even on large signals, the tuner will be biased back and the IF amplifier will be operating at an unnecessarily large gain and converter noise will be evident in the picture. If the ratio is too large, then no controlling bias will be applied to the tuner and it will be held at the clamping voltage and all control will be made in the IF amplifiers. This can cause severe overloading of the IF amplifier.

VERTICAL DEFLECTION CIRCUITS

Vertical sync. pulses from the sync. separator are used to synchronise the blocking oscillator, T3 and the triode portion of the valve. "Height" is adjusted by varying the DC potential fed to the blocking oscillator anode and "Vertical Hold" is adjusted by varying the time constant of the blocking oscillator grid circuit. The sawtooth waveform from the blocking oscillator is applied to the grid of the output amplifier and a sawtooth current waveform appears in the vertical output transformer. In 15-valve receivers, a feedback voltage is developed across R97, R98, from the current in the deflection coils. This voltage is stepped up to the input grid of the vertical output valve. A potentiometer, RV9, is provided for adjustment of linearity. In the Blumlein circuit, employed in the 13-valve receivers, two linearity controls are provided to achieve good, overall vertical linearity.

HORIZONTAL OSCILLATOR AND AUTOMATIC PHASE CONTROL

Automatic frequency and phase control is obtained by means of a DC controlled, cathode coupled, stabilised multivibrator controlled by a germanium diode phase discriminator.

Sync. pulses from the sync, separator are fed into the centre of the discriminator and a sawtooth waveform from the multivibrator output is fed across the diode loads.

Since the negative going sync. pulse is fed to the diode cathodes and the diodes are effectively in parallel, then the discriminator output will be zero volts. The sawtooth is not of sufficient amplitude to cause the diodes to conduct, due to the bias caused by the coupling capacitor, so that the DC component of the sawtooth (average AC) is zero volts. Neither the sawtooth nor the sync. pulses can cause a bias voltage to be developed across the discriminator but that part of the sawtooth that occurs at the instant of the sync. pulses will have an affect on the bias voltage produced. If the sync. pulse occurs in the centre of the sawtooth, then the output is zero volts, and if it occurs before the retrace passes through its zero axis then the oscillator is running slow and the output voltage will be negative. The reverse will be the case if the oscillator is running fast.

The frequency of the Horizontal Multivibrator is controlled by the DC output of the discriminator. If the output voltage of the discriminator is positive it causes the cathode voltage to rise, lengthening the discharge time of the coupling capacitor to the second triode and slows down the firing rate of the multivibrator. A negative output from the discriminator will have the reverse effect.

HORIZONTAL DEFLECTION CIRCUITS

The horizontal driver valve produces a negative pulse output which is timed to cut off the horizontal output valve at the end of a scan. When cut off sharply, the magnetic field that has been established in the horizontal output transformer during the scan collapses and the oscillatory circuit comprised of the transformer inductance and stray capacitances tends to "ring." However, after one-half cycle of oscillation the damping diode starts to conduct. During the "flyback" time the magnetic energy has established itself in the reverse direction, and the picture tube spot has returned to the left-hand side of the screen.

When the damping diode conducts it permits current to flow at a controlled rate through part of the transformer. This current, passed by the auto-transformer into the deflection coils, forms the initial part of the horizontal scan. As the damper ceases to conduct the line output valve takes over and supplies the necessary current to

The receiver is shipped from the factory with

complete the scan, at which point a further negative pulse on the grid starts the cycle over again.

During flyback a high voltage pulse is produced at the anode of the EHT rectifier, which is peak-rectified and then smoothed by the capacitance between inner and outer bulb coatings of the picture tube, and supplies EHT of approximately 16,000 volts.

Energy recovered by the damping diode produces a boosted HT voltage of 830 volts, which is divided down to 560 volts for supplying the G2 electrode voltage of the picture tube.

In 15-valve receivers, T7 is the focus transformer. The sawtooth scanning current in its primary winding produces, in the secondary, a large parabolic voltage waveform which is fed direct to the focus electrode of the CRT and the cold end of the secondary is connected to a suitable voltage to give good overall focus.

REMOTE CONTROL

(15-value receivers only).

By plugging into socket SKT2 the five-pin socket in the rear of the chassis, volume and contrast can be controlled from the Remote Control Unit.

The remote volume control RV6A adds a variable resistance across the supply voltage of the sound limiter. Since this control can only *reduce* volume, the main volume control should be set for the maximum volume desired.

The remote contrast control RV4A feeds a variable DC voltage into the contrast control chain R57, R58. It varies contrast either side of a middle (normal) value which is set by the main contrast control.

Note that connection of the Remote Control Unit does not affect operation of the normal receiver volume and contrast controls.

A facility is provided on the socket for use of a remote hearing aid.

INSTALLATION

PICTURE SHIFT

Small shifts in position of picture may occur due to the effect of the earth's magnetic field in different locations. The picture may be re-centred by rotating the two shift magnets on the tube neck behind the deflection yoke.

Rotate the centring magnet assembly to shift the picture in the required direction, and move one of the magnets with respect to the other to change the strength of the field and hence the amount of picture shift.

PICTURE TILT

If the picture is not square with the edges of the mask, the deflection coils should be rotated until the picture is squared up. It may be

the picture tube installed and all controls preadjusted for normal operation. It should only be necessary to ensure that the mains tapping is correctly adjusted for the mains voltage existing in the particular area and a suitable aerial connected to the aerial input terminals. In very strong signal areas it may be necessary to use an attenuator in the aerial lead to avoid overloading the receiver. The various operating controls should be checked for proper operation, and their use demonstrated to the purchaser as described in the installation manual. It is necessary to remove the back of the cabinet to gain access to the mains adjustment. necessary after this operation to centre the picture by means of the shift magnets.

A.G.C.

The A.G.C. control is normally preset in the factory but if it is necessary to adjust it at any time the procedure is to turn the control to the maximum anti-clockwise position, then observing the picture, advance the control until the noise or "snow" in the picture is no longer reduced. The receiver should then be checked on all channels to ensure that no overloading is evident, which may be due to the control being adjusted too far in a clockwise position, and that the minimum noise condition has been achieved for all signals.

FUSES

Three fuses are provided, one in the mains circuit and two in the HT circuits. Ensure that they are replaced with similar types.

NOISE INVERTER

The cathode bias of the noise inverter can be adjusted on installation (if necessary) by means of a pre-set control on the rear panel of the receiver. Tune to the strongest signal and turn contrast control to maximum. Turn the Noise Inverter control in an anti-clockwise direction until the picture tends to go out of lock. Turn control slightly clockwise so that picture returns to lock. Check that receiver locks quickly on all channels when changing channels.

SERVICING

The vertical chassis of this receiver has been especially designed to make servicing as easy as possible. All valves, test points and major components are accessible to the serviceman when the cabinet back is removed. All other components may be serviced by swinging the chassis out so that all of the receiver is accessible.

To do this, remove the EHT lead from its support; slacken off the screws on the righthand side of the chassis and remove the clamp. Note that one of these screws is intended as a factory transit screw only, and need not necessarily be replaced. Remove screw from below mains transformer. This is also a transit screw and need not necessarily be replaced. The chassis, pivoting about the left-hand side, may be swung out to an angle of approximately 40 degrees.

In order to swing it out further than this, disconnect the yoke and EHT anode lead.

To secure the chassis, reconnect any disconnected leads and replace the screw below the power transformer before the retaining clamp and its screws are replaced.

If the repair or replacement cannot be made without removing the chassis from the cabinet, the receiver can be withdrawn as a complete working unit and can be operated as such on the workshop bench.

DISMANTLING

Chassis Types M4, M6, P1 and P6

REMOVAL OF CHASSIS ASSEMBLY

Remove the bottom screws securing the back cover of the receiver to the cabinet. Ease the back cover down until the top edge is free of the cabinet groove. Withdraw the cover straight back over the picture tube neck.

WARNING: Be careful not to drop the cover on to the neck of the picture tube when the bottom screws are removed.

Pull off the four small knobs from the front of the receiver. Undo the grub screws and remove the brass collar on the contrast spindle. Set the Channel Selector knob to channel 5. Flex the black plastic cover back with a screwdriver at the channel 7 position. Undo the securing screw at this position and pull off both the Channel Selector and the Fine Tuning knobs.

Remove two bolts that secure the base board to the cabinet shelf.

Remove two bolts that secure the top of the picture tube clamp bracket to the cabinet.

Slacken off two screws on the antenna bracket. Slide the bracket toward the rear of the cabinet in its guide grooves and remove the bracket from the cabinet.

The chassis and picture tube may then be withdrawn from the cabinet.

Chassis Types M3, M5, M7, P2, P5, P7 and P8

REMOVAL OF CABINET

Remove the cabinet back by undoing the nine securing screws (two are underneath).

Disconnect the tuner lead from the antenna bracket by withdrawing the lead pins from the antenna socket.

Disconnect the speaker leads.

Pull off the four small knobs from the front of the receiver. Undo the grub screw on the contrast spindle and remove the brass collar. Set the Channel Selector knob to channel 5. Flex the black plastic cover back with a screwdriver at the channel 7 position and undo the securing screw. Pull off both the Channel Selector and Fine Tuning knobs.

Remove the two screws that secure the top of the picture tube clamp to the cabinet.

Remove six screws that secure the cabinet to the wooden base board.

Remove the cabinet.

REMOVAL OF PICTURE TUBE SCREEN

Once the cabinet has been removed, the front screen may be removed by undoing the four corner screws.

The covering material of the cabinet is washable vinyl, and if necessary may be cleaned with a damp cloth. The clear, protective picture tube screen and plastic moulded front may also be cleaned with a soft, damp cloth, but avoid the use of chemical and abrasive cleaners. Do not use any preparation normally sold as window cleaning or furniture polishing agents, as these generally prove harmful to those materials.

If the plastic becomes stratched or accidentally marked by a fly-spray, the marks may be removed by polishing gently with a soft cloth moistened with Wattyl or Dupol Cabinet Burnisher. Very deep scratches should be removed first with Water Rubbing Compound and then finished off with Burnisher.

DEFLECTION YOKE

First remove the picture tube socket. Loosen the clamp-fixing screws on the rear of the yoke assembly, remove the yoke plug from the socket on the rear of the EHT assembly and slide yoke over neck of tube. When replacing yoke, do not use force and do not tighten clamping screws until set has been operated and picture is squared up.

REMOVAL OF PICTURE TUBE

Having removed the chassis and picture tube assembly, remove the spring which rests against the aquadag coating on the rear of the picture tube. Undo the top and bottom screws that secure the clamping ring to the picture tube on 21-inch receivers, or slacken the retaining nut on 23-inch receivers. Lift the tube out carefully by supporting it around the mounting ring.

N.B.: The picture tube should be carefully handled and never placed face-down on a bench. Always ensure that it is placed on a soft, clean surface, such as felt, so that the face does not become stratched. Whenever possible, keep tubes in the original manufacturers' carton.

HORIZONTAL LINEARITY

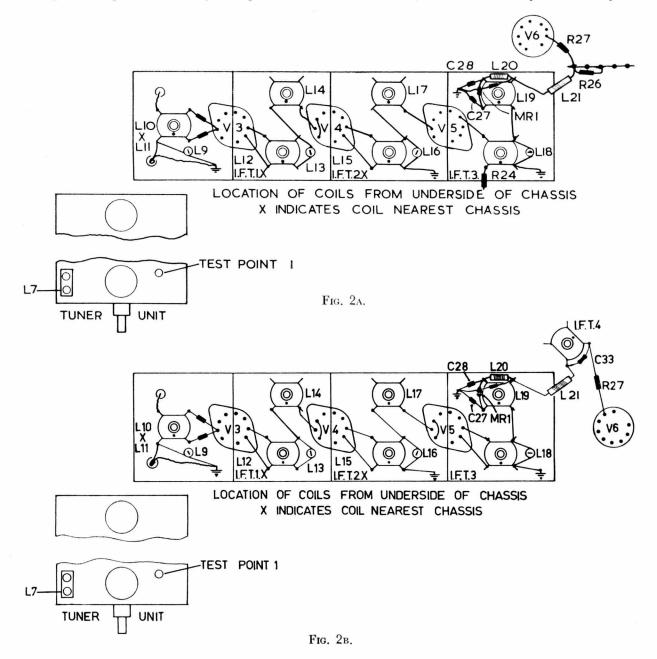
A typical multimeter employing a 100 ohm 1 mA fullscale deflection meter, when on 100 mA range, has a total resistance of 1 ohm. If such a meter is connected from pin 8 of the Horizontal Output valve to earth, *i.e.*, across the 1 ohm metering resistor, it will indicate half of the current flowing in the cathode of the valve. The Horizontal Linearity control should be adjusted to reduce this current to a minimum.

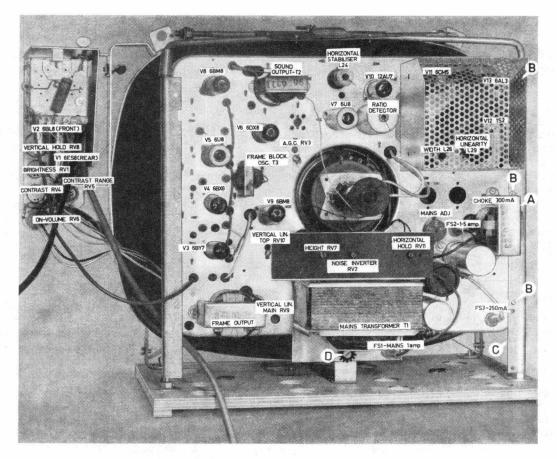
HORIZONTAL HOLD

Disconnect sync. pulses by removing the sync. separator valve, in 15-valve receivers, or earthing the grid test point of the sync. separator in 13-valve receivers. Short circuit the Horizontal Stabilising coil. Adjust the Horizontal Hold so that the picture "floats" or locks weakly. Remove the short across the stabilising coil and adjust the core of this coil so that the picture again floats or locks weakly. Replace the sync. separator valve or remove the short from the grid test point, whichever is applicable.

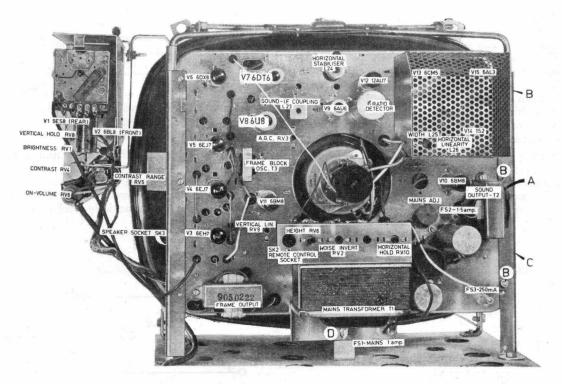
CONTRAST RANGE

Turn the Contrast Control to its maximum clockwise position and adjust the Contrast Range control to give sync. tips at 190 volts at the Video anode, read on a DC coupled oscilloscope.





REAR VIEW — 13-VALVE CHASSIS

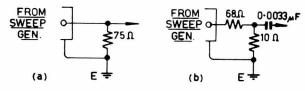


REAR VIEW — 15-VALVE CHASSIS

ALIGNMENT

VISION IF

To align the vision IF a sweep generator and a marker generator, both covering the range 28.5 to 38.5 Mc/s are required, together with a display unit. The marker generator may be a signal generator and the display unit a C.R.O. These instruments should be interconnected as described in the instructions supplied with the sweep generator. The sweep generator should be terminated with a resistor equal to its output impedance and connected to the receiver as shown in Figure 3.



E-EARTH OF CIRCUIT INTO WHICH SIGNAL IS BEING INJECTED.

FIG. 3.

Coils L9, L13, L16, L18 adjust the bandwidth of the coupling circuits and are adjusted and sealed in the factory. It should only be necessary to adjust these on realignment if IF transformers or coils have been replaced in the circuit.

Before commencing alignment, remove slugs from L11, L12 and L15, and screw the slugs of IFT1 and IFT2 to set flush with the chassis.

Connect a bias supply of -6 volts across the IF A.G.C. Connect the display unit across R27. Throughout the alignment the display unit should be adjusted to present a reasonable amplitude display from a signal 2.5 volts peak-to-peak, and the output from the IF strip should be maintained at that level by varying the output from the sweep generator.

Because of the high gain of the 15-valve receiver, care should be taken to ensure that all components replaced are on short leads and are placed in exactly the same position as the original part. Care must also be taken to prevent feedback

The following equipment is necessary to carry out this procedure:

- (i) A C.W. Oscillator accurately tuned to 5.5 Mc/s by a crystal controlled reference.
- (ii) A 20,000 ohm / volt meter (Model 8AVO or similar type).
- (iii) A DC V.T.V.M.

9

(iv) A peak-to-peak detector as shown.

in interconnecting leads of alignment equipment.

The following procedure must be followed step by step, and do not proceed to the next step until sure that each response has been accurately obtained.

- Connect the sweep via input as in 3a to pin 2 of V5. Adjust the slugs in IFT3 (slug in position nearest the chassis) and L19 (slug in position farthest from chassis) to achieve a response as shown in Fig. 5A.
- (2) Remove sweep from V5 pin 2 and connect through a terminating pad as in Fig. 3b to pin 2 of V4. Adjust the slugs of IFT2 (slug in position nearest chassis) and L17 to achieve the response shown in Fig. 5B.
- (3) Remove the sweep from V4 and connect, through the same terminating pad, to V3 pin 2. Adjust the response, with the slugs of IFT1 (slug nearest chassis) and L14 (slug farthest from chassis) to that shown in Fig. 5c.
- (4) Remove the sweep from V3 and connect to the tuner through Test Point 1, using the same terminating pad. Adjust L10 (with slug nearest chassis) and L7 to produce final response as shown by Fig. 5D.

Note: The correct final response will be obtained only if each stage is accurately aligned. It may therefore be necessary to slightly readjust coils other than L11 and L7, on the final alignment, to produce the response shown in Fig. 5n, exactly.

- (5) Insert slugs with retaining rubber, into L11, L12 and L15. Set L11 to 31.375 Mc/s., L12 to 29.875 Mc/s., and L15 to 38.375 Mc/s. (Fig. 5E).
- (6) Check overall response and adjust if necessary. Also, check stability by removing the bias and adjusting the input accordingly. The response should remain substantially unchanged.

SOUND I.F. ALIGNMENT

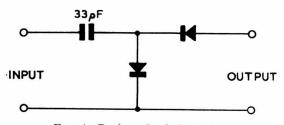
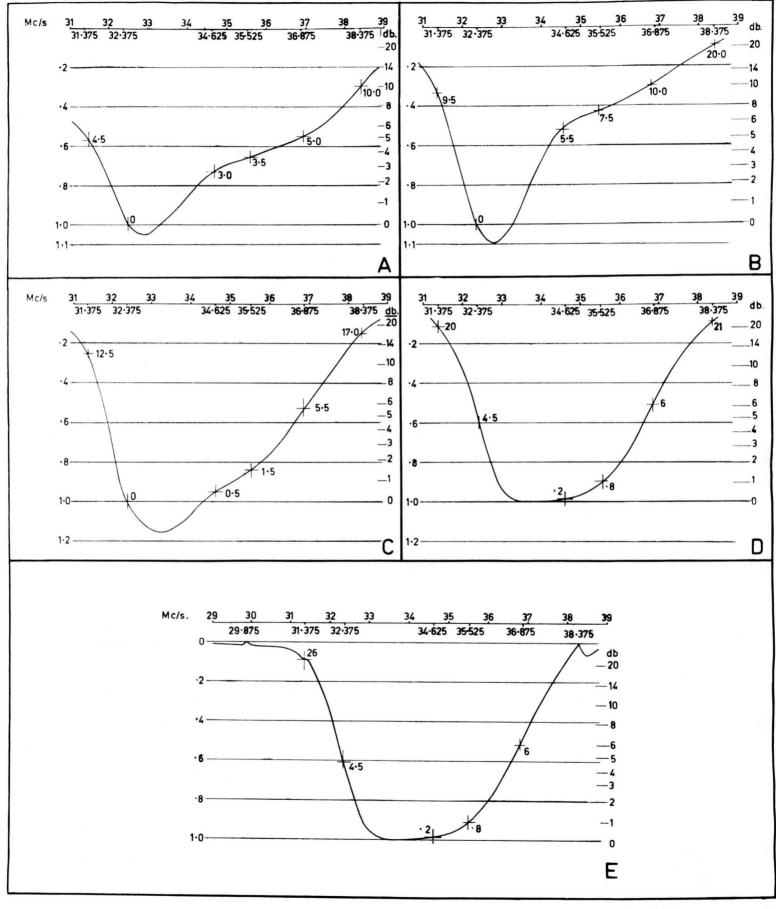


FIG. 4: Peak-to-Peak Detector.



5.5 MC/s. NULL TRAP (IFT4)

IFT4 is a combined null trap and transformer, working at 5.5 Mc/s. When tuned in the factory, both primary and secondary cores are tuned together to give a zero output at 5.5 Mc/s. at the video grid, and a maximum transfer to the intercarrier amplifier. This can only be done accurately with a sweep oscillator and a suitable display having a high gain at 5.5 Mc/s. Once set, however, it should not need retuning unless quite large circuit alterations have been made.

Should it be necessary to retune IFT4, the following precedure should be adopted:

- Inject 5.5 Mc/s. at approximately 100 mV between the junction L20 and L21 and earth (disconnecting the grid peaking choke, L20).
- (2) Connect the input of the peak-to-peak detector illustrated to CRT cathode, pin 7. Connect the output of the peakto-peak detector to a 20,000 ohm/volt meter on the 50 micro-amp range.
- (3) Adjust primary core of IFT4 (nearest chassis) to give zero reading on meter. If IFT4 is replaced, it is necessary to adjust both cores to give a zero reading at 5.5 Mc/s.

Physical differences in the bulbs of various single-face 23-inch picture tubes impose some restrictions on the type of tube which should be used for replacement purposes. Generally speaking, the 23-inch chassis can be divided into two groups, which can be identified by the shape of the chassis support rods in the region where they cross one another. Some rods are straight and others are bent in this region to provide a greater distance between chassis and picture tube clamping band.

Chassis	Replacement Tube Type
M3 M4 M5 M6 M7	AW53-88 or 21DAP4.

- (4) Withdraw both cores from former. Screw in primary core (nearest chassis) to give a minimum reading.
- (5) Screw in secondary core until meter reading rises slightly and then adjust primary core until a new minimum is obtained.
- (6) Repeat adjustment of primary and secondary cores until meter reads zero.

ANODE COUPLING COIL (L24)

(15-value receivers only)

With the oscillators connected as in (1) above and the VTVM connected across R67, adjust the core of L24 so that a maximum negative reading is obtained on the VTVM.

RATIO DETECTOR TRANSFORMER (IFT5)

Connect oscillator as in (1) above. Adjust secondary core (nearest chassis) so that a positive or negative reading is obtained on a DC VTVM between the junction of the diode loads and earth. Adjust primary core (top of coil) so that this reading shows a maximum. Then adjust secondary core so that this reading is zero volts. This procedure may be used on a signal from a transmission.

PICTURE TUBE REPLACEMENT

In chassis with straight rods only picture tubes with a "K" moulded on to the glass bulb are suitable. These types are AW59-30, 23WP4, 23MP4-K, and 2351B.

In chassis with bent rods, any of the 23-inch single-face tubes may be used.

A tabulation of the chassis and picture tube variations is given below:

Chassis	Replacement TubeType
P1 (early production) P2 P8	AW59-30, 23WP4, 23MP4-K, 2351B.
P1 (late production) P5 P6 P7	AW59-30, 23WP4, 23MP4-K, 2351B, 2354B. AW59-90, 23MP4-J,

MODELS M4, M5, M7, P6, P7 and P8

RESISTORS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
$\substack{\textbf{REF.}\\ \textbf{R13}\\ \textbf{R14}\\ \textbf{R15}\\ \textbf{R14}\\ \textbf{R15}\\ \textbf{R17}\\ \textbf{R19}\\ \textbf{R221}\\ \textbf{R224}\\ \textbf{R226}\\ \textbf{R228}\\ \textbf{R229}\\ \textbf{R332}\\ \textbf{R334}\\ \textbf{R356}\\ \textbf{R378}\\ \textbf{R356}\\ \textbf{R378}\\ \textbf{R356}\\ \textbf{R423}\\ \textbf{R445}\\ \textbf{R4456}\\ \textbf{R7445}\\ \textbf{R7456}\\ \textbf{R7551}\\ \textbf{R551}\\ \textbf{R5551}\\ \textbf{R555}\\ \textbf{R555}\\ \textbf{S555}\\ \textbf{R555}\\ \textbf$	$\begin{array}{c} \textbf{PART No.} \\ \hline \\ $	DESCRIPTION 2.2K ohms $\frac{1}{2}$ W \pm 10% 2.2K ohms $\frac{1}{2}$ W \pm 10% 56 ohms $\frac{1}{2}$ W Morganite \pm 10% 100 ohms $\frac{1}{2}$ W Morganite \pm 10% 2.2K ohms 1 W \pm 20% 1K ohms $\frac{1}{2}$ W Morganite \pm 10% 150 ohms $\frac{1}{2}$ W Morganite \pm 10% 2.2K ohms 1 W \pm 20% 2.2K ohms 1 W \pm 20% 2.2K ohms 1 W \pm 20% 2.2K ohms 1 W \pm 10% 100 ohms $\frac{1}{2}$ W Morganite \pm 10% 2.2K ohms 1 W \pm 10% 1.2K ohms $\frac{1}{2}$ W \pm 10% 1.8K ohms 2 W \pm 10% 12K ohms $\frac{1}{2}$ W \pm 10% 12K ohms 2 W \pm 10% 12K ohms 3 W \pm 10% 12K ohms 4 W \pm 10% 12K	REF. R58a R59 R60 R61 R62 R63 R64 R65 R66 R67 R68 R69 R70 R71 R72 R72a R74 R74 R74 R74 R75 R76 R77 R78 R78 R80 R82 R83 R84 R85 R86 R87 R88 R89 R90 R91 R92 R95 R98 R98 R99 R100	PART No. 740-0082 740-0112 740-0112 740-0752 742-0152 742-0152 740-0622 740-0622 740-0622 740-0622 740-0622 740-0622 740-0622 742-0272 742-0272 742-0252 740-0653 742-0162 742-0182 742-0182 742-0182 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0512 742-0182 740-01852 740-01852 740-0142 740-0142 740-0112 740-0112 740-0122 740-0572 742-062 742-0	10K ohms $\frac{1}{2}$ W \pm 10% 27K ohms $\frac{1}{2}$ W \pm 10% 27K ohms $\frac{1}{2}$ W \pm 10% 27K ohms $\frac{1}{2}$ W \pm 10% 56K ohms $\frac{1}{2}$ W \pm 10% 330K ohms 1 W \pm 10% 220K ohms 1 W \pm 10% 220K ohms 1 W \pm 10% 220K ohms 1 W \pm 10% 470K ohms $\frac{1}{2}$ W \pm 20% 470K ohms $\frac{1}{2}$ W \pm 10% 580 ohms 5 W WW \pm 10% 270 ohms 1 W Morganite \pm 10% 1.5M ohms 1 W \pm 10% 1.5M ohms 1 W \pm 10% 1.5M ohms 1 W \pm 10% 1.5K ohms $\frac{1}{2}$ W \pm 10% 1.5K ohms $\frac{1}{2}$ W \pm 10% 1.5K ohms $\frac{1}{2}$ W \pm 10% 2.2K ohms 1 W \pm 10% 2.2K ohms 1 W \pm 10% 2.2K ohms 1 W \pm 10% 2.70 ohms $\frac{1}{2}$ W \pm 10% 470K ohms $\frac{1}{2}$ W \pm 10% 560K ohms $\frac{1}{2}$ W \pm 10% 560K ohms $\frac{1}{2}$ W \pm 10% 560K ohms $\frac{1}{2}$ W \pm 10% 15K ohms 1 W \pm 10% 15K ohms 1 W \pm 10% 56K ohms $\frac{1}{2}$ W \pm 10% 15K ohms 1 W \pm 10% 56K ohms $\frac{1}{2}$ W \pm 10% 15K ohms 1 W \pm 10% 56K ohms $\frac{1}{2}$ W \pm 10% </td
R56 R57 R58	$\begin{array}{r} 740-0122\\ 742-0392\\ 740-0082\end{array}$	47K ohms $\frac{1}{2}$ W + 10% 47K ohms 1 W + 20% 10K ohms $\frac{1}{2}$ W + 10%	R103 R104	$\begin{array}{c} 908-0381 \\ 740-0492 \\ 749-0302 \end{array}$	1.5 ohms Wire Resistor 1.5M ohms ½ W ± 20% 150K ohms 2 W ± 10%

CAPACITORS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
C14 C15	273-0591 271-0031	68 pFd MS silver mica ± 2½% .0033 mFd Ceramic disc G.M.V.	C37	269-0511	80 mFd Electrolytic 300V Wkg., with C38
C16 C17	271-0031 271-0031 271-0031	.0033 mFd Ceramic disc G.M.V. .0033 mFd Ceramic disc G.M.V.	C38	269 - 0511	40 mFd Electrolytic 300V Wkg., with C37
C18	271-0031	.0033 mFd Ceramic disc G.M.V.	C39	271-0031	.0033 mFd Ceramic disc G.M.V.
C19	280-1651	.0018 mFd Styroseal 600V + 5%	C40	269-0611	4 mFd Electrolytic 300V Wkg.
C20	273-0591	68 pFd MS silver mica + 21%	C41	279-4661	.022 mFd Paper Tubular 400V
C21	271 - 0031	.0033 mFd Ceramic disc G.M.V.			+ 10%
C22	280-1651	.0018 mFd Styroseal 600V + 5%	C42	280-1791	220 pFd Styroseal 600V + 10%
C23	273-0591	68 pFd MS silver mica + 21%	C43	273-0921	68 pFd MS Silver Mica + 10%
C24	271-0031	.0033 mFd Ceramic disc G.M.V.	C44	279-1161	.22 mFd Paper Tubular 200V
C25	271-0031	.0033 mFd Ceramic disc G.M.V.			+ 20%
C26	271-0031	.0033 mFd Ceramic disc G.M.V.	C45	279-0281	1 :nFd Metallised Paper 200V
C27	271 - 0471	6.8 pFd Ceramic Disc NPO			$\pm 25\%$
		± 1 pFd	C46	271-0231	68 pFd Ceramic Disc N750 3KV
C28	271 - 0131	8.2 pFd Ceramic Tube NPO	Transmission 1		± 10%
		± 1 pFd	C47	279-1161	.22 mFd Paper Tubular 200V
C29	271-0181	15 pFd Ceramic Tube NPO	100.000		± 20%
		$\pm \frac{1}{2}$ pFd	C48	271-0351	33 pFd Ceramic Tube NPO ± 5%
C30	271 - 0221	2.2 pFd Ceramic Bead NPO	C49	271-0031	.0033 mFd Ceramic Disc, G.M.V.
C31	271 - 0311	± 1 pFd	C50	273 - 0561	10 pFd Type IF ± 10%
G 6 6	000 0011	27 pFd Ceramic Tube NPO ± 5%	C51	273-0031	100 pFd type IF ± 5%
C32	269-0211	8 mFd Electrolytic ET2D 300V	C52	280-1501	100 pFd Styroseal 600V \pm 5%
000	070 1701	Wkg.	C53	280-1501	100 pFd Styroseal 600V ± 5%
C33	279-4701	.047 mFd Paper Tubular 400V	C54	269-0371	10 mFd Electrolytic ETIB 25V
C34	050 4541	$\pm 10\%$	OFF	070 1011	Wkg. .015 mFd Paper Tubular 400V
034	279 - 4541	.0022 mFd Paper Tubular 400V	C55	279-1641	+ 20%
C34a	279-4661	$\pm 10\%$	C56	280-1851	$\pm 20\%$ 680 pFd Styroseal 600V + 10%
0348	279-4661	.022 mFd Paper Tubular 400V		279-4001	.01 mFd Paper Tubular 200V
C35	269-0521	$\pm 10\%$	C57	219-4001	+ 10%
030	209-0921	100 mFd Electrolytic, Insulated, 150V Wkg.	C58	279-4621	.01 mFd Paper Tubular 200V
C36	269-0531	100 mFd Electrolytic, Uninsulated.	098	219-4021	+ 10%
0.50	200-0001	150V Wkg.	C59	269-0061	16 mFd Electrolytic 300V Wkg.
		100V WAR.	0.09	200-0001	to mru meenoryde soor wrg.

MODELS M4, M5, M7, P6, P7 and P8

CAPACITORS—continued

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
C60	269-0371	10 mFd Electrolytic 25V Wkg.	C77	279-0561	.5 mFd Metallised Paper 200V
C62	279-2281	.22 mFd Paper Tubular 600V			± 25%
		$\pm 20\%$	C78	280 - 1751	100 pFd Styroseal 600V ± 10%
C63	279-1121	.1 mFd Paper Tubular 2VV	C79	280 - 0061	.0033 mFd Styroseal $200V \pm 5\%$
		$\pm 20\%$	C80	280 - 1861	820 pFd Styroseal 600V \pm 10
C64	279-4701	.047 mFd Paper Tubular 400V	C81	280 - 1751	100 pFd Styroseal 600V \pm 10%
~		$\pm 10\%$	C82	280 - 1781	180 pFd Styroseal 600V \pm 10%
C64a	269 - 0371	10 mFd Electrolytic 25V Wkg.	C83	279 - 4581	.0017 mFd Paper Tubular 400V
C65	279 - 4661	.022 Paper Tubular 400V \pm 10%			± 10%
C66	279-4661	.022 Paper Tubular 400V \pm 10%	C84	271-0481	82 pFd Ceramic Tubular ± 5%
C67	279-4701	047 Paper Tubular 400V ± 10%	C85	271 - 0341	150 pFd Ceramic Disc 5KV
C68	269-0481	24 mFd Electrolytic 300V Wkg.,			$N750 \pm 10\%$
0.00		with C69	C85a	279-5161	.022 mFd Paper Tubular, 600V
C69	269-0481	24 mFd Electrolytic 300V Wkg.,			$\pm 20\%$
		with C68	C86	279-1781	.22 mFd Paper Tubular 400V
C70	269-0361	100 mFd Electrolytic 25V Wkg.	and the second second		$\pm 20\%$
C71	279-4681	.033 mFd Paper Tubular 400V	C87	279 - 5771	22 mFd Paper Tubular 1KV
		Wkg. $\pm 10\%$	100000000		$\pm 10\%$
C72	273 - 1061	75 pFd Silver Mica ± 5%	C88	279-5701	.047 mFd Paper Tubular 1KV
C73	273-1031	47 pFd MS silver mica 🛨 1 pFd	1.2000		$\pm 10\%$
C74	280-1891	.0015 mFd Styroseal $600\overline{V} \pm 10\%$	C89	279-4161	.22 mFd Paper Tubular 200V
C75	280-1001	.001 mFd Styroseal 400V 🛨 10%			± 10%
C76	279-1581	.0047 mFd Paper Tubular 400V ± 20%	C90	271-0431	470 pFd Ceramic Disc, K2000, 3KV ± 20%
			1		

COILS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
L9	259-0672	Tuner Coupling Trim	L19	259-0931	Detector Input
L10)	259-0941	1st IF Grid	L20	259-0953	Grid Peaking Choke
111 5	259-0941	31.375 Mc/s Trap	L21	259-0953	Grid Peaking Choke
L12	Part of	CONTRACTOR Designment of Contractors	L22	259-1081	Peaking Choke-Shunt
	906-0232	29.875 Mc/s Trap (On IFT1)	L23	259-1091	Peaking Choke-Series
.13	259-1061	1st IF Coupling	L24	259-0992	Horizontal Stabilising Coil
.14	259-1061	2nd IF Grid	L25	259-0044	Coil. Anti-Parasitic
15	Part of		L26	259-0902	Width Control
	906-0281	38.375 Mc/s Trap	L27	259-0044	Coil. Anti-Parasitic
.16	259-0672	2nd IF Coupling	L28	259-0044	Coil. Anti-Parasitic
.17	259-1061	3rd IF Grid	L29	259-0922	Horizontal Linearity Contro
18	259-0672	3rd IF Coupling	1120	200 0021	

POTENTIOMETERS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
RV1	677-0641	500K Cure A type Q—Brightness	RV7	677-0343	250K Curve A type EC-Height
RV2	677-0171	25K Curve A type EC — Noise Inverter	RV8	677-0841	100K Curve A type Q — Vertical Hold
RV3 RV4	$677 - 0421 \\ 677 - 0601$	500K Curve A type ES—A.G.C. 25K Curve A type Q—Contrast	RV9	677-0343	250K Curve A type ES Vertical Linearity Main
$\mathbf{RV5}$	677-0611	50K Curve C type ES-Contrast Range	RV10	677-0343	250K Curve A type ES - Vertical Linearity Top
RV6	677-0621	1M ohms tapped 500K Curve A, with D.P.P.P. Switch-Volume	RV11	677-0631	50K Curve A type P.T.UHori- zontal Hold

TRANSFORMERS

				1			
REF.	PART No.		DESCRIPTION	REF.	PART No	•	DESCRIPTION
IFT1 906-0232 IFT2 906-0281 IFT3 906-0253 IFT4 906-0261 IFT5 906-0101		IF Vision IF Vision IF Vision Sound Take off and 5.5 Mc/s Trap Ratio Tarnsformer		T1 T2 T3 T4 T5	$\begin{array}{r} 904\text{-}0253\\ 905\text{-}0312\\ 908\text{-}0461\\ 905\text{-}0226\\ 908\text{-}0381\end{array}$	Power Transformer Audio Output Blocking Oscillator Vertical Output Horizontal Output	
			VA	LVES			
REF.	PART No.		DESCRIPTION	REF.	PART No		DESCRIPTION
V1 V2 V3 V4	932-1161 932-0501 932-0881 932-0521	6ES8 6BL8 6BY7 6BX6	RF Amplifier Frequency Changer 1st IF Amplifier 2nd IF Amplifier	V7 V8 V9	932-1101 932-0511 932-0511	6U8 6BM8 6BM8	Limiter and A.G.C. Audio Drive & Output Blocking Oscillator an Vertical Output
V5 V6	932-10321 932-1101 932-1081	6U8 6DX8	3rd IF Amplifier and Noise Inverter Video Amplifier and Sync. Separator	V10 V11 V12 V13	932-0481 932-0531 932-0771 932-1151	12AU7 6CM5 1S2 6AL3	Horizontal Multivibrat Horizontal Output EHT Rectifier Damping Diode

MODELS M4, M5, M7, P6, P7 and P8

MISCELLANEOUS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
CH1	232-0124	Choke 300 mA	FS3	431-0031	Fuse, 250 mA.
MR1	932-0971	Diode 0A80	Yoke	259-1171	21-inch Yoke, with Thermistor
MR2	932-1071	Diode 0A210			RT1 for M4, M5, M7.
MR3	932-1071	Diode 0A210	Yoke	259-1101	23-inch Yoke, with Thermistor
MR3a	932-0791	Diode 0A81			RT1, for P6, P7, P8.
MR4	932-0991	Diode M3	CRT	932-1121	AW53-88 M4, M5, M7
MR5)	932-0601	Diodes. 2 - 0A79		932-1121	21DAP4
MR6 (332-0601	Diddes, $2 = 0A79$		932-1262	AW59-30
MR7	932-0791	Diode 0A81		932-1262	23MP4
MR8	932-0791	Diode 0A81		932-1262	23MP4-K
RT2	752-0021	Thermistor CZ4		932-1261	23MP4-J } P6, P7, P8
VDR1	750-0281	Voltage Dependent Resistor, type		932-1261	AW59-90
		E298GD/A260		932-1262	2351B
FS1	431-0071	Fuse, 1 amp.		932-1261	2354B
FS2	431-0081	Fuse, 1.5 amp.		Refer to Te:	xt before replacing CRT.

MODELS M3, M6, P1, P2 and P5

RESISTORS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
R13	740-0032	2.2K ohms $\frac{1}{2}$ W \pm 10%	R67	740-0082	$10K \text{ ohms } \frac{1}{2}W \pm 10\%$
R14	740-0032 740-0983	2.2K ohms $\frac{1}{2}$ W \pm 10% 22 ohms $\frac{1}{2}$ W Morganite \pm 10%	R68 R69	749-0052 740-0032	47 K ohms 2 W \pm 20%
R15 R16	740-0983	100 ohms $\frac{1}{2}$ W Morganite \pm 10%	R70	740-0032	2.2K ohms $\frac{1}{2}$ W \pm 10% 10K ohms $\frac{1}{2}$ W \pm 10%
R17	742-0712	2.2K ohms 1 W $+$ 20%	R70a	740-0082	10K ohms $\frac{1}{2} \text{W} \pm 10\%$
R18	740-0682	680 ohms $\frac{1}{2}$ W \pm 10%	R71	740-0112	27K ohms $\frac{1}{2} \text{W} + 10\%$
R20	740-0983	22 ohms 1 W Morganite + 10%	R72	740-0112	27K ohms 1 W + 10%
R21	740-0273	150 ohms 1 W Morganite + 10%	R73	740-0122	47K ohms 1 W + 10%
R22	742-0712	2.2K ohms 1 W \pm 20%	$\mathbf{R74}$	740-0702	56K ohms 1 W + 10%
R23	740-0032	2.2K ohms $\frac{1}{2}$ W \pm 10% 150 ohms $\frac{1}{2}$ W Morganite \pm 10%	R75	742-0492	68 K ohms 1 W \pm 10%
R24	740-0273		R76	742-0132	220 K ohms 1 W $\pm 10\%$
R25 R26	742 - 0322 742 - 0322	1K ohms 1 W \pm 20% 1K ohms 1 W \pm 20%	R77 R78	$740-0062 \\ 740-1052$	3.9 ohms $\frac{1}{2}$ W \pm 10%
R27	740-0042	2.7K ohms $\frac{1}{2}$ W \pm 10%	R79	740-1062	330K ohms $\frac{1}{2}$ W \pm 20% 680K ohms $\frac{1}{2}$ W \pm 20%
R28	740-0572	1K ohms $\frac{1}{2}$ W + 10%	R80	750-0332	680 ohms 5 W W.W. $\pm 10\%$
R29	749-0242	1.8K ohms 2 W + 10%	R81	742-0823	270 ohms 1 W Morganite \pm 10%
R30	749-0242	1.8K ohms 2 W + 10%	R 82	742-0112	100K ohms 1 W \pm 10%
R31	740-0773	39 ohms 1 W Morganite ± 10%	R83	740-0112	27K ohms 1 W ± 10%
R32	742-0162	390K ohms 1 W \pm 10%	R84	740-0082	$10K \text{ ohms } \frac{1}{2} \text{ W } \pm 10\%$
R33	740-0122	47K ohms $\frac{1}{2}$ W \pm 10%	R85	740-0082	$10 \mathrm{K} \mathrm{ohms} \frac{1}{2} \mathrm{W} \pm 10\%$
R34	740-0622	470K ohms 1 W + 20%	R86	740-0122	47K ohms 1 W ± 10%
R35 R35a	740-0732 740-0142	12K ohms $\frac{1}{2}$ W \pm 10% 100K ohms $\frac{1}{2}$ W \pm 10%	R87 R88	740-0082	10K ohms 1 W + 10%
R36	750-0291	250 ohms 5 W W.W. $+$ 5%	R89	742-0172 740-0082	470K ohms $1 \text{ W} \pm 10\%$
R37	749-0142	$1 \text{K ohms } 2 \text{ W} \pm 20\%$	R90	740-0232	10K ohms $\frac{1}{2}$ W \pm 10% 39K ohms $\frac{1}{2}$ W \pm 10%
R38	749-0252	$12 \text{K} \text{ ohms } 2 \text{ W} \pm 10\%$	R91	742-0022	4.7K ohms 1 W \pm 10%
R39	740-0142	100K ohms 1 W + 10%	R92	740-0202	2.2M ohms $\frac{1}{2}$ W \pm 10%
R40	740-0162	220K ohms 1 W + 10%	R93	740-0582	47K ohms 1 W + 20%
R41	740-0162	220K ohms 1 W ± 10%	R94	740-0302	1.8K ohms 1 W + 10%
R42	740-0242	33K ohms 1 W ± 10%	R95	749-0142	1K ohms 2 W + 20%
R43	740-0162	220K ohms $\frac{1}{2}$ W \pm 10%	R96	742-0823	270 ohms 1 W Morganite ± 10%
R44 R45	$740-0822 \\ 740-0122$	33K ohms $\frac{1}{2}$ W \pm 20% 47K ohms $\frac{1}{2}$ W \pm 10%	R97 R98	740 - 1043 740 - 1043	27 ohms 1 W Morganite ± 10%
R45 R46	740-0112	27 K ohms $\frac{1}{2}$ W + 10%	R99	740-1043	27 ohms $\frac{1}{2}$ W Morganite \pm 10% 270K ohms $\frac{1}{2}$ W \pm 10%
R47	742-0772	3.9M ohms 1 W \pm 10%	R100	740-0392	330 K ohms $\frac{1}{2}$ W $\frac{1}{4}$ 10%
R48	740-0202	$2.2M$ ohms $\frac{1}{2}W + 10\%$	R101	740-0182	470K ohms $\frac{1}{2}$ W $\frac{1}{2}$ 10%
R49	742-0122	150K ohms 1 W + 10%	R102	740-0142	100K ohms $\frac{1}{2}$ W + 10%
R50	749-0232	27K ohms 2 W + 10%	R103	740-0102	22K ohms 1 W ± 10%
R51	742-0772	3.9M ohms 1 W ± 10%	R104	742-0042	15K ohms 1 W + 10%
R52	742-0772	3.9M ohms 1 W ± 10%	R105	742-0472	1.8K ohms 1 W ± 10%
R53	742-0732	1.8 M ohms 1 W \pm 10%	R106	740-0702	56K ohms $\frac{1}{2}$ W \pm 10%
R54 R55	742 - 0182 740 - 0782	680K ohms 1 W \pm 10%	R107 R108	740-0112	27K ohms 1 W ± 10%
R56	740-0702	120K ohms $\frac{1}{2}$ W \pm 10% 56K ohms $\frac{1}{2}$ W \pm 10%	R108 R109	$740-0122 \\742-0172$	47K ohms $\frac{1}{2}$ W \pm 10%
R57	740-0162	220K ohms $\frac{1}{2} \times \frac{10\%}{10\%}$	R109	740-0572	470K ohms $1 \text{ W} \pm 10\%$ 1K ohms $\frac{1}{2} \text{ W} \pm 20\%$
R58	742-0982	$1.2M$ ohms $1 W \pm 10\%$	R111	750-0182	2.7K ohms 5 $W W.W. + 10\%$
R59	740-0782	120K ohms 1 W ± 10%	R112	746-0242	$1 \text{ ohm } \frac{1}{2} \text{ W W.W. + } 10\%$
R60	749-0232	27 K ohms 2 W + 10%	R113	742-0262	2.7K ohms 1 W + 10%
R61	740-0082	10K ohms 1 W ± 10% 150 ohms 1 W Morganite ± 10%	R114	742-0492	68K ohms 1 W + 10%
R62	740-0273	150 ohms 1 W Morganite ± 10%	R115	742-0112	100K ohms 1 W ± 10%
R63	740-0572	1K ohms $\frac{1}{2}$ W \pm 10% 470K ohms $\frac{1}{2}$ W \pm 20%	R116	740-0492	1.5M ohms 1 W ± 20%
R64	740-0622	$470K$ ohms $\frac{1}{2}W \pm 20\%$	R117	Part of	1.5 Jan Wiles Desister
R65 R66	740-0142 740-0122	100K ohms $\frac{1}{2} \mathbb{W} + 10\%$ 47K ohms $\frac{1}{2} \mathbb{W} + 10\%$		908-0381	1.5 ohm Wire Resistor.
1000	140-0122				

CAPACITORS

DESCRIPTION
.0022 mFd Paper Tubular 400V
$a_{00} \pm \frac{10\%}{10\%}$
.022 mFd Paper Tubular 400V
\pm 10% 100 mFd Electrolytic (Insulated)
150V Wkg.
100 mFd Electrolytic (Uninsula-
ted) 150V Wkg.
80 mFd Electrolytic 300V Wkg.
(With C43)
40 mFd Electrolytic 300 VWkg.
(With C42)
4 mFd Electrolytic 300V Wkg.
.022 mFd Paper Tubular 400V
$\pm 10\%$
220 pFd Styroseal 600V ± 10%
68 pFd MS Silver Mica ± 10%
.22 mFd Paper Tubular 200V
± 20%
.1 mFd Paper Tubular 400V
± 20% 1 mFd Metallised Paper 200V
+ 25%
68 pFd Ceramic Disc N750 3kV
+ 10%
.0033 mFd Ceramic Disc G.M.V.
.22 mFd Paper Tubular 200V
+ 20%

MODELS M3, M6, P1, P2 and P5

CAPACITORS — continued

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
C52	526-2962	1.5 mFd Lead Capacitor	C78	279-4721	.068 mFd Paper Tubular 400V
C53	279-4541	.0022 Paper Tubular 400V ± 10%			$\pm 10\%$
C54	271-0031	.0033 mFd Ceramic Disc G.M.V.	C79	279-4621	.01 mFd Paper Tubular 400V
C55	279-1621	.01 mFd Paper Tubular 400V	G 0.0		$\pm 10\%$
0.00		$\pm 20\%$	C80	269-0481	24 mFd Electrolytic 300V Wkg.
C56	271-0471	6.8 pFd Ceramic Disc NPO	0.01	050 1101	(With C82) .22 mFd Paper Tubular 200V
C57	271-0491	± 1 pFd 10 pFd Ceramic Disc NPO + 10%	C81	279-1161	+ 10%
C58	280-1651	.0018 Styroseal 600V + 5%	C82	269-0481	\pm 10% 24 mFd Electrolytic 300V Wkg.
C59	273-0561	10 pFd Type I/F $+$ 10%	082	269-0481	(With C80)
C60	271-0031	.0033 mFd Ceramic Disc G.M.V.	C83	269-0361	100 mFd Electrolytic 25V Wkg.
C61	273-0331	100 pFd Type I/F + 5%	C84	273-1051	82 pFd MS Silver Mica + 10%
C62	280-1501	100 pFd Styroseal $600V + 5\%$	C85	273-1031	47 pFd MS Silver Mica + 1 pFd
C63	280-1501	100 pFd Styroseal 600V \pm 5%	C86	280-1891	.0015 mFd Styroseal 600V + 10%
C64	269-0371	10 mFd Electrolytic 25V Wkg.	C87	280-1001	.001 mFd Styroseal 400V + 10%
C65	280-1851	680 pFd Styroseal 600V Wkg.	C88	279-1581	.0047 mFd Paper Tubular 400V
000	200 1001	+ 10%	000	210 2002	+ 20%
C66	279-4001	.01 mFd Faper Tubular 200V	C89	279-0561	.5 mFd Metallised Paper 200V
		+ 10%			+ 25%
C66a	280-1791	220 pFd Styroseal 600V + 10%	C90	280-1751	100 pFd Styroseal 600V ± 10%
		(P1, P2, P5 only)	C91	280-0061	.0033 mFd Styroseal 200V ± 5%
C67	279-4001	.01 mfd Paper Tubular 200V	C92	280-1861	820 pFd Styroseal 600V \pm 10%
		$\pm 10\%$	C93	271-0481	82 pFd Ceramic Tube NPO \pm 5%
C68	269-0371	10 mFd Electrolytic 25V Wkg.	C94	280-1781	180 pFd Styroseal $600V \pm 10\%$
C69	279-1641	.01 mFd Paper Tubular 400V	C95	279-4581	.0047 mFd Paper Tubular 400V
050	000 0001	$\pm 20\%$			$\pm \frac{10\%}{10\%}$
C70 C71	269-0061	16 mFd Electrolytic 300V Wkg.	C96	271-0481	82 pFd Ceramic Tubular NPO
C72	269-0221 279-1661	25 mFd Electrolytic 25V Wkg.	007	051 0041	\pm 5% 150 pFd Ceramic Disc 5 kV N750
072	279-1661	.022 mFd Paper Tubular 400V	C97	271-0341	+ 10%
C73	279-0281	± 20% 1 mFd Metallised Paper 200V	C98	279-2281	.22 mFd Paper Tubular 600V
013	279-0281	+ 25%	0.98	279-2281	+ 20%
C74	279-4541	.0022 mFd Paper Tubular 400V	C99	279-5771	.22 mFd Paper Tubular 1 kV
011	210-1011	+ 10%	000	215-0111	+ 10%
C75	279-1701	$\frac{10\%}{10}$.047 mFd Paper Tubular 400V	C100	279-5701	.047 mFd Paper Tubular 1 kV
		+ 20%	2100		+ 10%
C76	279-4541	.0022 mFd Paper Tubular 400V	C101	279-4161	.22 mFd Paper Tubular 200V
		+ 10%			+ 10%
C77	279-4541	.0022 mFd Paper Tubular 400V	C102	271-0431	470 pFd Ceramic Disc K2000
		$\pm 10\%$			3 kV + 20%

COILS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
L9 L10) L11 } L12 L13 L14 L15 L16 L17 L18	259-0672 259-0941 Part of 906-0232 259-0672 259-1061 Part of 906-0281 259-0672 259-1061 259-0672	Tuner Coupling Trim 1st IF Grid 31.375 Mc/s Trap 29.875 Mc/s Trap (On IFT1) 1st IF Coupling 2nd IF Grid 38.375 Mc/s Trap (On IFT2) 2nd IF Coupling 3rd IF Grid 3rd IF Coupling	L19 L20 L21 L22 L23 L24 L25 L25a L26 L26a L26a L27 L28	$\begin{array}{c} 259 - 0931 \\ 259 - 0953 \\ 259 - 0953 \\ 259 - 1081 \\ 259 - 1081 \\ 259 - 1091 \\ 259 - 0992 \\ 259 - 0044 \\ 259 - 0992 \\ 259 - 0044 \\ 259 - 0902 \\ 259 - 0044 \\ 259 - 0922 \\ 259 - 0444 \end{array}$	Detector Input Grid Peaking Choke Grid Peaking Choke Peaking Choke—Shunt Peaking Choke—Series Sound IF Anode Horizontal Stabilising Coil Coil—Anti-Parasitic Width Coil Coil—Anti-Parasitic Horizontal Linearity Coil Coil—Anti-Parasitic

POTENTIOMETERS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
RV1	677-0641	500K Curve A Type Q-Brightness	RV6a	677-0801	250K Curve A Type PTU - Tone
$\mathbf{RV2}$	677-0171	25K Curve A Type EC — Noise Inverter	RV7	677-0931	(P1, P2, P5 only) 50K Curve A Type EC—Height
-					
RV3 RV4	677 - 0911 677 - 0601	1M Curve A Type EC—A.G.C. 25K Curve A Type Q—Contrast	RV8	677-0641	500K Curve A Type Q — Vertical Hold
RV5	677-0611	50K Curve C Type EC-Contrast Range	RV9	677-0511	10K Curve A Type EC - Vertical Linearity
RV6	677-0621	1M Tapped 500K Curve A with D.P.P.P.—Switch Volume	RV10	677-0631	50K Curve A Type PTU - Hori- zontal Hold.

TRANSFORMERS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
IFT1	906-0232	IF Vision	T 2	905-0303	Audio Output
IFT2	906-0281	IF Vision	T 3	908-0321	Blocking Oscillator
IFT3	906-0253	IF Vision	T 4	905-0226	Vertical Output
IFT4	906-0261	Sound Take off and 5.5 Mc/s Trap	T 5	908-0352	Vertical Feedback
IFT5	906-0101	Ratio Transformer	T 6	908-0381	Horizontal Output
T1	904-0253	Power Transformer	T 7	908-0391	Focus Transformer

MODELS M3, M6, P1, P2 and P5

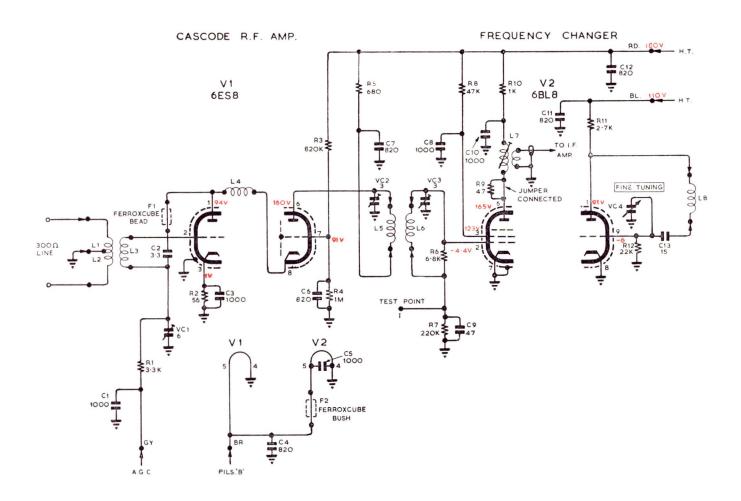
VALVES

REF.	PART No.		DESCRIPTION	REF.	PART N	ίο.	DESCRIPTION
V1	932-1161	6ES8	RF Amplifier	V9	932-0441	6AU6	Limiter
V 2 V 3	$932 - 0501 \\ 632 - 1211$	6BL8 6EH7	Frequency Changer 1st IF Amplifier	V10	932-0511	6BM8	Audio Driver and Output
V4 V5	932-1221 932-1221	6EJ7 6EJ7	2nd IF Amplifier 3rd IF Amplifier	V11	932-0511	6BM8	Blocking Oscillator and Vertical Output
V 6	932-1081	6DX8	Video Amplifier and Noise Inverter	V12 V13	932-0481 932-0531	12AU7 6CM5	Horizontal Multivibrator Horizontal Output
V7	932-1231	6DT6	Sync. Separator	V14	932-0771	182	EHT Rectifier
V 8	932-1101	6U8	Sound Amplifier and A.G.C.	V 15	932-1151	6AL3	Damping Diode

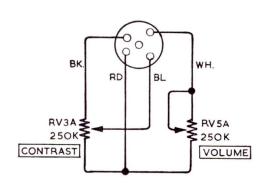
MISCELLANEOUS

REF.	PART No.	DESCRIPTION	REF.	PART No.	DESCRIPTION
CH1	232-0124	Choke 300 mA	FS3	431-0031	Fuse, 250 mA.
MR1	932-0971	Diode 0A80	Yoke	259-1011	21-inch Yoke for M3, M6
MR2	932-1071	Diode 0A210		259-1051	23-inch Yoke for P1, P2, P5
MR3	932-1071	Diode 0A210	CRT	932-1211	AW53-88 M3. M6
AR4	932-0991	Diode M3		932-1121	21DAP4 (M.S. M.S.
MR5) MR6 (932-0601	Diodes (2-0A79)		932 - 1262 932 - 1262	AW59-30 23WP4
AR7	932-0791	Diode 0A81		932-1262	23MP4-K
IR8	932-0791	Diode 0A81		932-1261	23MP4-J P1, P2, P5
RT1	752-0021	Thermister CZ4		932-1261	AW59-90
VDR1	750-0281	Voltage Dependent Resistor		932-1262	2351B
		E298GD/A260		932-1261	2354B
TS1	431-0071	Fuse, 1 Amp.			,
FS2	431-0081	Fuse, 1.5 Amp.		Refer to tex	t before replacing CRT.

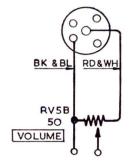
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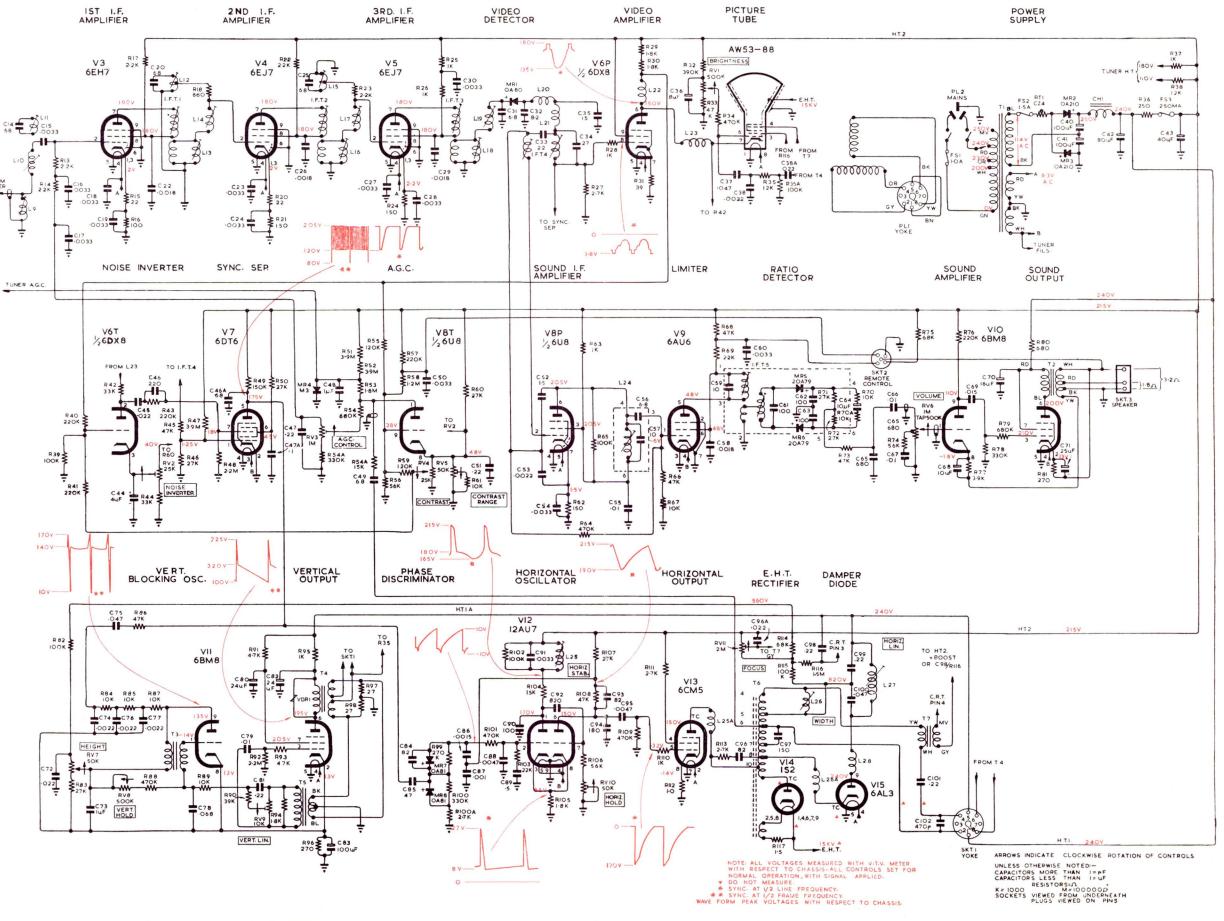


REMOTE CONTROL



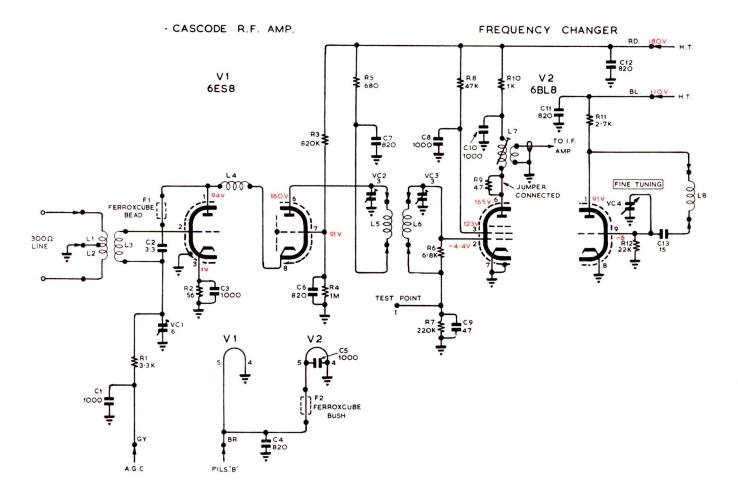
HEARING AID CONTROL UNER





"H·M·V" CHASSIS TYPE M3 & M6

"H·M·V" CHASSIS TYPE PI, P2 & P5



HEARING AID

CONTROL

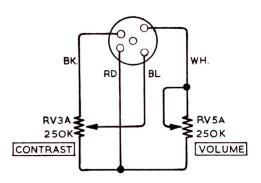
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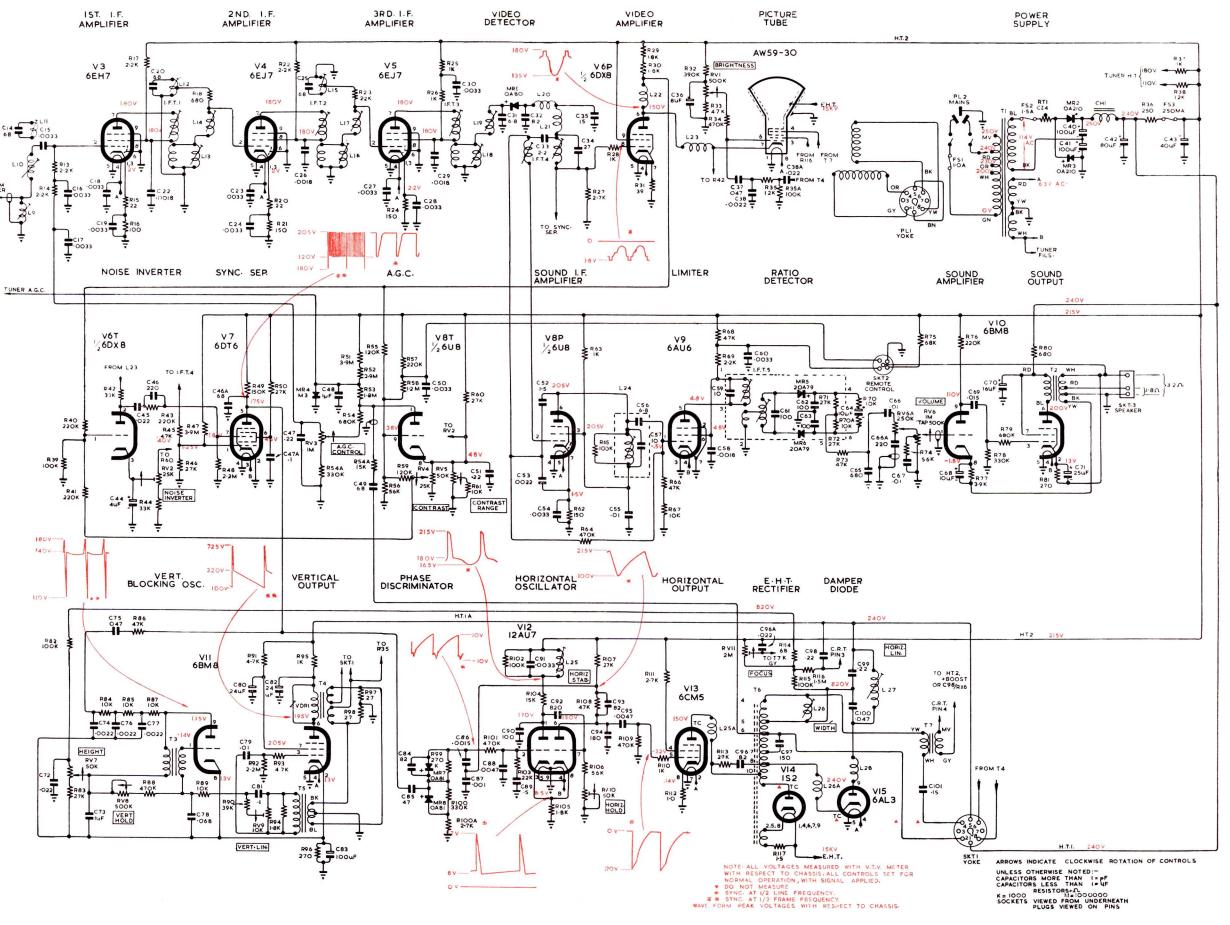
BK & BL RD&WH

RV58

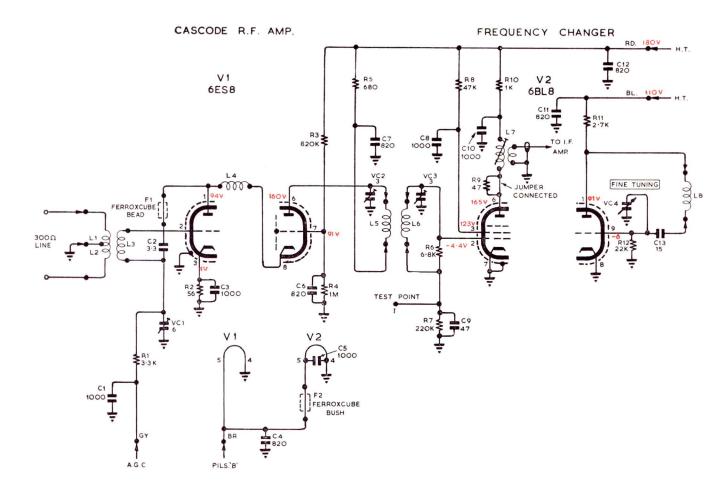
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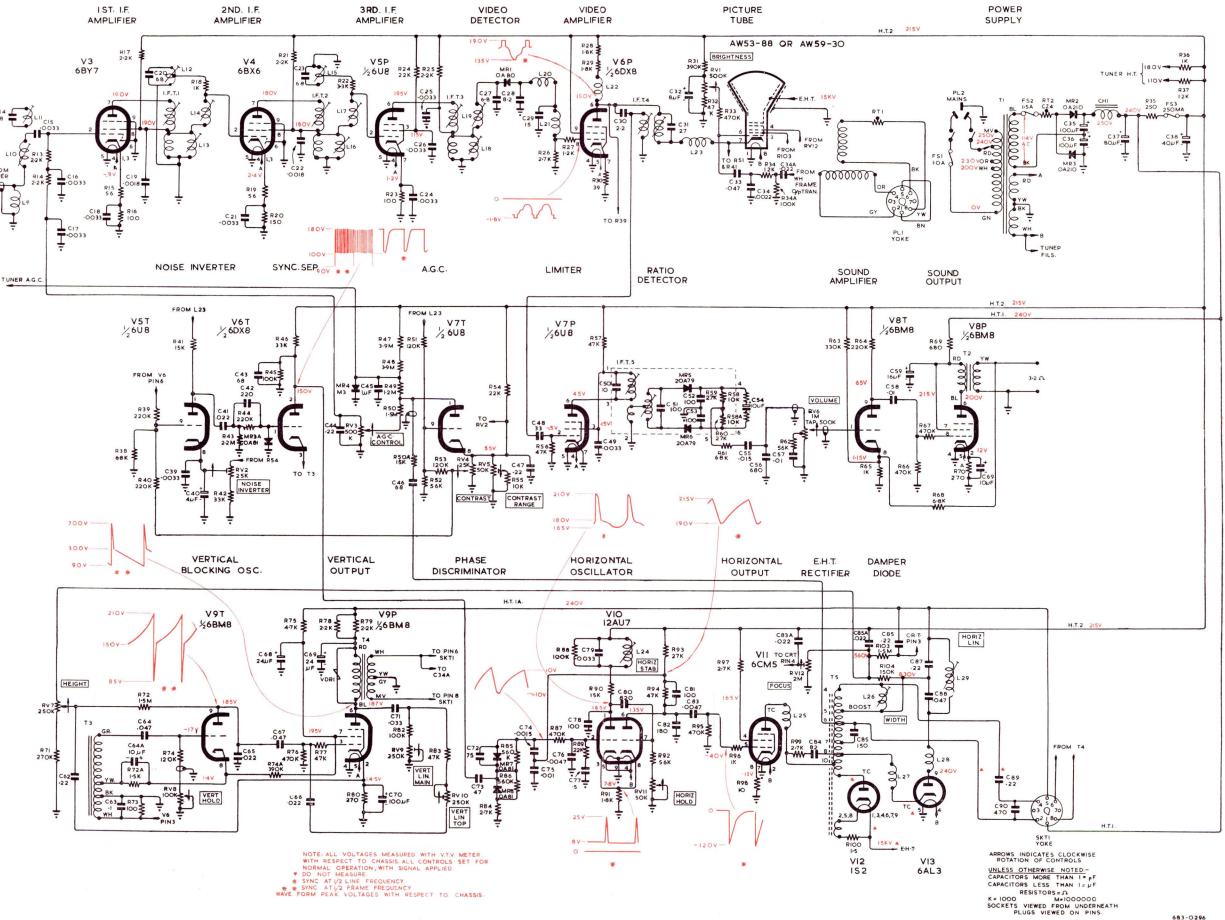
REMOTE CONTROL





UNER





"H·M·V" CHASSIS TYPE M4, M5, M7, P6, P7 & P8

683-0296