

WORKSHOP



MODEL bwd-509A
5" SINGLE BEAM
OSCILLOSCOPE

B.W.D. ELECTRONICS PTY. LTD.
331-333 BURKE ROAD, GARDINER
VICTORIA, AUSTRALIA

INSTRUMENT HANDBOOK

Applicable to Serial No. **9971**.....

MODEL bwd-509A

5" SINGLE BEAM

OSCILLOSCOPE

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INSTRUMENT HANDBOOK

MODEL bwd-509A

1. GENERAL: The major requirements in a modern oscilloscope are; simplicity of operation, compactness, reliability and of course a first class performance. How well these features have been combined in the bwd 509A oscilloscope becomes self evident as soon as a signal is displayed. Direct reading controls enable voltages below 50mV to over 500V from DC to over 5MHz to be measured against time scales from 1 second to 200 nano Seconds - a range of 5 million to 1.
- 1.1 The bug-bear of a stability control needed on many oscilloscopes to set up the trigger condition is completely eliminated in this model either externally or internally, instead diode clamps lock the all solid state time base circuit in a sensitive ready state which is completely unaffected by input voltage changes from 170 to 265V or 85 to 132V - a truly stable circuit.
- 1.2 The time base and trigger circuit also incorporates other new techniques to ensure rock steady triggering. The gated AUTOMatic time base produces a bright reference line at all time base speeds - even at 1 μ Sec/cm with no signal present, and is teamed with a wide band trigger circuit which is preset for optimum sensitivity. As it does not contain the AUTO circuit it is not subject to annoying beats and jitter which often occur when input signal and the AUTO free run rate interact particularly when displaying low level pulse waveforms.
- 1.3 To ensure that readings of voltage or time are within specification irrespective of variations of local power lines, both the amplifier and time base is compensated to accommodate $\pm 10\%$ line change. Tappings on power transformer permit larger variations to be accepted to suit local supply conditions.

2. SPECIFICATION

- 2.1 C.R.T.
- | | |
|------------------|---|
| <u>Type</u> | 5" flat faced, type 5U1F incorporating DC coupled Beam Blanking to the control grid. |
| <u>Phosphor</u> | P1, normally supplied, P7 available as an option. |
| <u>EHT</u> | 1.6KV |
| <u>Graticule</u> | 8 x 10cm graticule with 2mm subdivisions on X & Y centreline and green filter (orange filter for P7 Phosphor option.) |

2.2 VERTICAL AMPLIFIER

<u>Bandwidth</u>	DC or 2Hz (AC coupled) to 5MHz - 3db at all sensitivities referred to 4cm deflection at 1kHz.
<u>Sensitivity</u>	50, 100, 200, 500mV, 1, 2, 5, 10, 20, 50 & 100V per cm.
<u>Rise Time</u>	70 nano Seconds for 4cm deflection.
<u>Calibration</u>	Better than 5%
<u>Input Impedance</u>	1M Ω and approximately 43pF
<u>Max. Input Voltage</u>	400DC or 250AC or 400V AC and DC p-p combined.

2.3 TIME BASE

<u>Range</u>	1 μ Sec/cm to 100mSec/cm in 6 decade ranges with an uncalibrated vernier control between each step, extending range to 1Sec/cm.
<u>Calibration</u>	Better than 5% at all settings at X 1 magnification, 1 μ Sec to 10mSec/cm <10% at 100mSec/cm.
<u>Expansion</u>	X 1 to approx X 5 continuously variable. Maximum sweep speed 200nSec/cm at X 5 mag.
<u>Blanking</u>	Direct coupled to C.R.T. Grid.

2.4 TRIGGER

<u>Facilities</u>	2 switches and one switched potentiometer provide selection of following characteristics:
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INT or EXT
+ or -

AUTO or LEVEL SELECT

<u>Sensitivity</u>	Int. AUTO - 3Hz to 10MHz with 1cm deflection sine or square wave. 1Hz to 10MHz with 4cm deflection.
<u>Sensitivity</u>	Int. Level Select Range \pm 3cm deflection. Ext. AUTO 1V p-p 7Hz to 10MHz sine or square 5V p-p 1Hz to 10MHz sine or square Ext. Level Select range \pm 10V p-p min.
<u>Ext. Trigger Impedance</u>	100K Ω and 20pF approximately. Max input \pm 30V AC, DC, or AC and DC combined.

2.5 HORIZONTAL AMPLIFIER

<u>Bandwidth</u>	DC to 250kHz (-3db) at all sensitivities.
<u>Sensitivity</u>	500mV to 2.5V per cm approx. continuously variable.
<u>Input Impedance</u>	150K Ω and 15pF approx.
<u>Vertical to Horz. Ampl. Phase Shift</u>	Less than 5° from DC to 10kHz.

2.6 'Z' MODULATION

Rear socket input to C.R.T. grid, input time constant 0.01 MFD and 100K Ω . + 20V will modulate trace at normal intensity.

<u>Calibrate Waveform</u>	1V p-p sine wave 50Hz unstabilised
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2.7 POWER REQUIREMENTS

190 to 260V, 50 to 60Hz, approximately 30 watts.
95 to 130V 50 to 60Hz (110V option).

<u>Dimensions</u>	9" (23cm) high x 7" (18cm) wide x 16" (41cm) deep.
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<u>Weight</u>	Approximately 16 lbs. (7 $\frac{1}{2}$ kg)
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<u>Accessories</u>	Supplies with instrument
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1 Handbook, Circuit and Parts List.
1 Power Cord.

<u>Optional Accessories</u>	Refer Section 13 page 15 of this Handbook.
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3. FUNCTION OF CONTROLS

- 3.1 Front Panel controls are grouped for ease of use and are clearly designated. The functions of these controls are as described below:

<u>INTENSITY CONTROL</u>	Fully anti-clockwise, this control switches the instrument OFF, When rotated clockwise the instrument is switched ON and further rotation controls the trace intensity (brightness) from zero to max.
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<u>FOCUS</u>	Controls the sharpness of the trace. May require a slight re-adjustment over the full intensity control range.
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<u>HORZ. SHIFT</u>	(Red Knob) moves the trace horizontally on the C.R.T.
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<u>HORZ. GAIN</u>	(Grey Knob) when the Time Base is in use, this control varies the length of the trace from 10cms to approx. 50cms, providing X5 magnification. When an External Horizontal Input is used, the Horz. Gain varies the sensitivity from 500mV to 2.5V per cm approximately.
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FUNCTION OF CONTROLS (Cont'd.)

T.B. VERNIER

Varies the Time Base speed over a 12 - 1 range approx. to provide a continuously variable range in conjunction with the TIME/CM switch from 1Sec/cm to 1 μ Sec/cm. When the VERNIER control is turned and switched fully anti-clockwise it switches off the internal Time Base permitting an external signal to be fed into the HORZ INPUT socket.

TIME/CM (Time Base) Switch

When the Time Base Vernier control is fully clockwise in the CAL position the 6 time base speeds on this control will be accurate to within 5%. The switch speeds represent the fastest speed on each range; rotation of the Time Base Vernier Control anti-clockwise will reduce the selected speed over a range greater than X 12, e.g. on the 1mSec range the Vernier will vary the time base from 1mSec down to approx. 12mSec/cm when fully anti-clockwise.

+ - SLOPE SELECTION

Selects the positive (+) or negative (-) slope of the displayed signal or external trigger waveform to initiate the time base.

INT. EXT. SWITCH

Selects the trigger signal from either the displayed waveform, or an external waveform applied to the EXT trigger socket. Line frequency trigger is available by link connection from the 1V p-p CAL socket to EXT trigger socket and by switching to EXT.

VERTICAL SHIFT

Moves the trace up and down the C.R.T. face.

AUTO. TRIG. LEVEL

Fully anti-clockwise, and switched to the AUTO position, any signal greater than 0.5cm in amplitude will trigger the time base but with no input the time base operates automatically producing a base line, the automatic rate increases as the Time Base speed range increases. When the knob is switched away from the AUTO position it permits selection of the point on the displayed or externally coupled waveform over a range of ± 3 cms which will trigger the time base.

VOLTS/CM (Attenuator)

Switch adjusts the sensitivity of the Vertical Amplifier from 50mV (.05V) per cm to 100V per cm in a 1, 2, 5, 10 series of steps. Attenuator accuracy is 2% and the overall oscilloscope accuracy is within 5% on any step.

DC - AC SWITCH

In the DC position of this switch the amplifier is directly coupled from input to output. In the AC position a capacitor is placed in series with the input to block the DC component of a signal. The AC component also is attenuated - 3db at 2Hz approx.

3.2 TERMINALS AND SOCKETS FRONT PANEL

VERTICAL INPUT SOCKET

'E'

A positive input will cause the trace to move upwards; a negative input will cause the trace to move downwards. Black terminal, should be connected to the ground side of the signal being measured.

HORZ. INPUT

When the Time Base Vernier is turned anti-clockwise to "T.B. OFF" signals may be fed into the HORZ. INPUT socket to produce a horizontal display, Input is DC coupled. If sufficient DC is present on the signal to bias the trace off the screen a blocking capacitor must be placed in series with the input signal to remove the DC.

CAL. 1V p-p

A 1V p-p (approx) sine waveform is available to check the oscilloscope operation. T.B. calibration, or if linked to the EXT trigger input to provide line frequency triggering.

EXT. TRIGGER

When the Trigger Selection switch is in the EXT position, signals from 1 to 20V will trigger the time base. Full selection of amplitude over a range of $\pm 10V$ or AUTO with positive or negative selection is available.

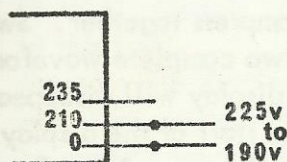
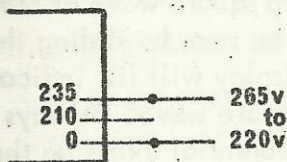
3.3 REAR PANEL

'Z' Modulation

A 20V p-p square wave or a sine wave of 6V RMS or greater will blank the trace. Positive going signals blank the trace. Negative signals brighten the trace.

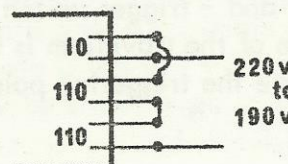
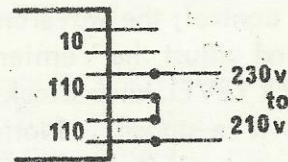
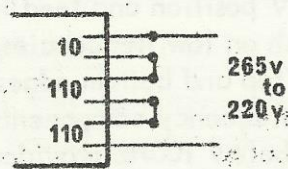
4. FIRST TIME OPERATION Check tapping on Power Transformer for correct connection for local supply mains. For operation from 220 to 260V use 235V tap. For operation from 190 to 220 use 215V tap. If instrument is fitted with universal primary for 110 to 230V operation connect as shown below.

230V TRANSFORMER

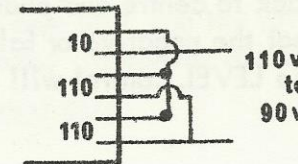
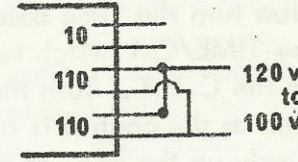
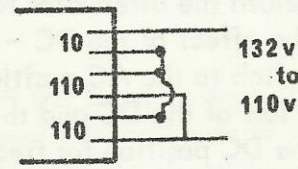


UNIVERSAL TRANSFORMER

200-240V CONNECTIONS



100-120V CONNECTIONS



4.1 Set the controls of Model bwd 509A as follows before switching on:

INTENSITY	Fully anti-clockwise
FOCUS	Centred
HORZ. SHIFT	Centred
HORZ. GAIN	Anti-clockwise
TIME BASE VERNIER	Clockwise - CAL
TIME/CM	10mSec
+ - SELECTOR	+
INT - EXT TRIGGER	INT
VERT SHIFT	Centred
TRIGGER LEVEL	Fully anti-clockwise - AUTO
VOLTS/CM	0.2V
DC - AC	DC

4.2 Connect instrument to power mains, check as above in 4.

Connect a link of wire from the CAL 1V p-p front panel socket to the vertical input socket on the L.H. side.

Switch on by rotating the INTENSITY control about $\frac{3}{4}$ of a turn. A display will appear after a few seconds.

5 cycles of the calibration waveform should be present on the C.R.T. Adjust the HORIZONTAL & VERTICAL SHIFT controls to centre the display and the FOCUS & INTENSITY for a sharp, bright image.

4.3 Now turn the VOLTS/CM switch to 0.1V and the display will expand over full screen height, turning the knob around to 0.5V, 1, 2, etc. will progressively reduce the height of the display, below 0.5cm amplitude the trace may start to lose stability.

4.4 Return the attenuator to the 0.2V position and feed in a 1V p-p square wave at 50Hz. The effect of the DC - AC switch on low frequencies can now be seen by sliding the switch to the AC position. The top and bottom edges of the display will tilt indicating a loss of the DC and the lowest frequency components in the square wave. Always use the DC position for frequencies below 100Hz provided no DC potential exists on the waveform sufficient to cause it to deflect off the C.R.T. Now turn the Time Base Vernier control; the waveforms will compress together. Switch the TIME/CM switch to 1mSec and adjust the Vernier to give two complete waveforms on the C.R.T. Turn the TRIGGER LEVEL knob clockwise; the display will disappear, then as the control is turned it will re-appear. Notice how the start of the display moves slowly up the edge of the waveform until it disappears again at the top. Bring the control back to centre and change the + and - trigger switch over to -; the display will change so that the negative or falling slope of the waveform is triggering the display. Rotation of the LEVEL control will again move the triggering point up or down the waveform.

- 4.5 Return the control to AUTO and adjust the TIME/CM switch to give 5 waveforms across the C.R.T. , then turn the HORZ. GAIN control fully clockwise, 1 waveform will expand to approx. 10cm; this illustrates the trace expansion facility. If the HORZ. SHIFT is turned, the trace can be tracked along to view any part of it from one end to the other.
- 4.6 To check the HORZ. INPUT, turn the Time Base Vernier to T.B. OFF (anti-clockwise.) Connect a lead from the CAL socket to the DC HORZ. INPUT socket directly above it. A horizontal line will appear whose length can be varied by the HORZ. GAIN control from approximately 2cm down to less than 3mm. The HORIZONTAL position of the trace can be set by the HORZ. SHIFT control.
- 4.7 'Z' Modulation - feed in a 1kHz oscillator to the RED rear panel socket marked 'Z' Mod. With an input of 6V RMS or $\pm 10V$ approx. and the trace at normal brightness level the base line will be intensity modulated. Positive going signals blank the trace whilst negative going signals brighten the trace.

5. MEASUREMENT OF DC (DIRECT) VOLTAGES

- 5.1 Set LEVEL CONTROL to AUTO. Switch the DC - AC switch to DC. For an initial test take a $1\frac{1}{2}V$ dry cell and set the attenuator to 0.5V. Connect the negative and to the 'E' terminal, set the trace to the centre of the graticule, touch a lead from positive end of the battery to the Red INPUT socket; the trace will move up 3cms, i.e. $3 \times 0.5V = 1.5V$. Now reverse the connections to the battery and note how the trace moves down 3cms. This illustrates how an oscilloscope can display positive and negative voltages or both simultaneously, e.g. when viewing a sine or square wave.

- 5.2 NOTE: The $1M\Omega$ input impedance of the oscilloscope must be taken into account when measuring high impedance points such as anode, grid or screen voltages of valve or the gate of F.E.T.'s working with high value loads. Where loading is critical a bwd high impedance probe, type P23, with an input of $10M\Omega$ shunted by 12pF should be used.

The DC input facility may be used to measure AC waveforms swinging about a DC voltage, as at the collector of a transistor or the anode of a valve, to check for bias settings or anode bottoming, etc. Maximum DC input should not exceed X10 input attenuator setting if it is required to re-centre the trace to view a signal superimposed on it.

6. MEASUREMENT OF AN AC (ALTERNATING) VOLTAGE

- 6.1 Set the DC - AC switch to AC and the Attenuator to 50V (if the input voltage is unknown). Connect a lead from the 'E' (Black) input terminal to the ground (earth) side of the signal to be measured, then connect a lead from the Red input socket to the signal source. (B.W.D. Models 112, 140A or 602 Sine Wave Oscillators are suitable for initial experiments in this test).

Increase the Vertical sensitivity by the VOLTS/CM switch until a display between 2 and, say, 8cm exists. Now adjust the Time Base switch and Vernier to enable the waveform to be readily seen. To measure the voltage of the displayed waveform, measure its overall height in centimetres by the calibrated graticule, then multiply this by the attenuator setting and the result is in Volts p-p e.g. if the display is 6cm high and the attenuator is set to 0.5V, then the amplitude is $6 \times 0.5 = 3V$ peak to peak. To convert to RMS voltage for sine waves, divide the 3V by 2.84, e.g. $\frac{3}{2.84} = 1.06V$ RMS

7. TIME and FREQUENCY MEASUREMENT

- 7.1 The frequency of the waveform can be found by turning the Time Base Vernier to CAL (clockwise) then switch the TIME/CM switch to a range where the signal can be clearly seen, e.g. if a waveform is 3cm long and the switch is on 100 μ Sec, then the duration of the waveform is $3 \times 100 \mu\text{Sec} = 300 \mu\text{Sec}$. The frequency can be determined by dividing 1 second, i.e. 1,000,000 μ Sec by the duration of the waveform -
- $$\frac{1,000,000}{300} = 3,333\text{Hz}$$

8. MEASUREMENTS WITH AN EXTERNAL HORIZONTAL INPUT

- 8.1 As the HORZ. INPUT is directly coupled, the C.R.T. display can be used for X - Y plotting over an 8 x 10cm area. Switch Time Base Vernier to T.B. OFF, centre display and adjust focus and intensity for a fine spot. Positive or negative voltages may now be applied to X and Y inputs and the result plotted on tracing paper placed over the C.R.T. or transferred to a ruled graph paper. AC signals will show phase displays or Lissajous figures. Less than 5° phase shift exists from DC to 10kHz between X and Y inputs.

9. CIRCUIT DESCRIPTION

- 9.1 VERTICAL AMPLIFIER. Signals applied to the red INPUT terminal are switched straight through to the attenuator in the DC position of S1 or via C1 in the AC position which blocks the DC component. Switch S2A - D attenuates the input signal in a 1, 2, 5, 10 sequence to an amplitude suitable for displaying on the C.R.T. Section S2A and B attenuates the signal in a 1, 10, 100, 1000 sequence every 3rd step. Section S2C and S2D steps the input down in the 1, 2, 5 sequence, therefore the two sections cascaded produce the 1, 2, 5, 10 attenuation steps. To maintain a constant AC to DC attenuation ratio the resistors are bypassed by capacitors adjusted so the the C x R value of the series arm is equal to the C x R shunt arm at each step. Constant input capacitance is maintained by C2 and C3 input shunt capacitors. R13 is an overload protection resistor bypassed by C15 for AC signal components. The Vertical Amplifier consists of V1A and B input differential Amplifier, Q1 & Q2 emitter followers inter-stage coupling driving Q3, Q4 and V2 output cascade C.R.T. driver. V1 incorporates voltage compensation in its cathode to stabilise against line changes. Gain is adjusted by RV1 shunt control located between V1A and B anodes. Vertical Shift is applied symmetrically to the anodes by R15 and R18 from RV2 front panel shift control. Q1 and Q2 reduce the capacitive loading on the input stage and provide a low impedance drive to Q3 and 4. H.F. compensation of the Amplifier is provided by capacitors C18 and 19; C16 and R29 correct the square wave response.

The C.R.T. Y plates are directly coupled to V2 anodes and internal trigger take off is from V2A anode via R32, C20 R102 divider.

- 9.2 **TRIGGER CIRCUIT** Internal or external trigger signals are selected by S3 and applied via C21 and Q5 phase splitter. S4A selects the + or - trigger signal whilst S4B by-passes Q5 emitter in the +ve position to increase H.F. amplification on +ve. Q6 and 7 form a Schmitt Trigger which generates a precise amplitude fast rise and fall pulse from any input signal large enough to trigger it. Assuming S5A is open the action is as follows: With Q6 conducting its collector will bottom and Q7 will be cut off by the voltage divider across R42, 101, 44, and RV9. A negative going input will cut off Q6, its collector will rise pulling Q7 into conduction producing a negative pulse across its collector load. Q6 and 7 have a common emitter load therefore current through Q7 will hold Q6 cut off until the input signal changes polarity and rises positively reversing the switching action. When S5A is opened the trigger point of the input waveform is no longer preset by R39 and 40 but can be pulled + or - by the potential on RV8 thus providing level selection of the trigger point on the input waveform. Trigger sensitivity is set in the AUTO position by RV9 sensitivity preset.
- 9.3 **TIME BASE** This circuit consists of Q9 and 10 bi-stable trigger, Q12 source follower driving Q13 Miller integrator and Q14 emitter follower output. Q8 is the AUTO gating transistor and Q11 the blanking generator. Diode D4 gates the Miller stage, D1 and 2 clamp Q9 and 10, D6 sets the trace length and D5 the starting level of the saw tooth waveform.
- The operation is as follows:
- With S5B closed (trigger Level Select in use).
- 9.4 Assuming Q9 is conducting, Q10 will be cut off, its collector will rise and D4 will conduct, pulling the gate of Q12 and hence the base of Q13 positive. The collector of Q13 will fall to approximately +3V pulling down Q14 base. At this point diode D6 connected into the emitter load of Q14 passes below zero and starts to conduct reducing the conduction of D4.
- 9.5 This stabilises in a quiescent state with the trace ready for a trigger input pulse from Q7 via C25. A negative pulse on Q9 base will cause its collector to rise taking Q10 base positive; this causes current to flow through Q10 through the emitter resistor R58, biasing Q9 off further creating a rapid cumulative action in which Q9 cuts off and Q10 saturates. D4 becomes reverse biased, Q12 is left with its gate at $\sim 1.5V$ approximately and connected through the timing resistor R63, 64 and 65 or RV14 as selected by S6C to a negative potential on RV10 which will endeavour to pull Q12 towards cut-off. However, the timing capacitors selected by S6D are effectively in circuit between the base and collector of the Miller transistor Q13 and will be charged up by the current through the timing resistor.
- 9.6 Q12 FET source follower presents a high impedance to the charging circuit enabling high value charging resistors to be utilised with small high stability timing capacitors. Q14 emitter follower provides a low output impedance to charge the timing capacitors and drive the output and gating circuits. As Q12 gate and Q13 base fall, Q13 collector rises and via Q14, R68 and C34 a charge is applied to the selected timing capacitor on S6D.

The result of this negative feedback is to linearise the charging rate of the timing capacitor and to produce a positive going sawtooth waveform at the collector of Q13 and via the DC coupling to Q14 where it appears at low impedance at the emitter. The sawtooth continues to rise until the potential at the junction of R70 and R71 reaches approx. 6V, D5 then conducts and charges C28 to 31 and C32 as selected by S6B. It also takes the base of Q9 positive to its emitter potential and continues positively until Q9 conducts causing its collector to fall, cutting off Q10 and at the same time transferring the emitter current from Q10 to Q9. D4 conducts pulling the gate of Q12 positively, Q13 collector falls, rapidly discharging the timing capacitor until Q14 emitter falls sufficiently to cause D6 to conduct to pull D4 back to a quiescent condition and stabilise the circuit condition ready for the next trigger pulse. This will initiate the next trace once the hold-off capacitor C28 - 32 have discharged through R49 and the base current of Q9 and D1 clamps the base of Q9 in its ready state.

9.7

Auto time base operation is obtained when S5B is opened. During the sweep time Q10 is conducting its collector is negative to ground so Q8 whose base is connected via R46 to Q10 collector conducts and clamps R47 at ground potential and discharges the selected capacitor C27 to C30. During the return trace period Q10 ceases to conduct, its collector rises and turns off Q8. The selected Auto capacitor can now charge through the divider R47, 48 and 49. The junction of R47 and 48 drops and if no trigger signal is present to initiate the circuit it will continue negatively until D1 becomes reversed biased pulling down the diode clamp divider causing Q9 to become forward biased thus initiating the time base to produce one sweep. The action is repeated until a trigger pulse is generated to lock the time base.

9.8

Q11 blanking generator is a PNP transistor operating between ground and the -54 rail. It is also switched into conduction when Q10 conducts. R57 is part of Q10 load and current through it pulls Q11 into saturation during the sweep time. During the return trace R56 and the current through R84, 85 Q17, RV7 and R101 in the C.R.T. circuit pulls Q11 collector down until D3 conducts and clamps it at -54V. C.R.T. blanking is obtained by operating Q17 as a high impedance constant current stage with its base held at +9V to the -1390V rail. RV7 and R101 set the current through the stage so that the voltage drops across R84 and 85 equals the -ve EHT Voltage. The collector impedance of Q17 is approx. $4M\Omega$ so $2/3$ of the blanking signal is applied to the C.R.T. grid. H.F. bypass is via C57.

9.9

HORIZONTAL AMPLIFIER V3a and b together with Q15 and 16 form a balanced cascade amplifier in which the time base is applied to Q15 base and external horizontal signals to Q16 base. The time base sawtooth waveform is attenuated by R69, C59, R72 divider and superimposed on a horizontal shift voltage taken off RV12A before applying it to Q15 base. Horizontal gain is controlled by RV13X1 preset and RV12B HORZ GAIN control. R75 sets the X 5 magnification limit.

9.10 Input to the HORZ INPUT socket passes via R82 limiter to R81, C58 R80 attenuator to Q16 base. When the time base is not operating S7B opens the horizontal input grounding circuit and S7A open the emitter lead of Q9 and Q10, disconnecting the Time Base and unblanking the C.R.T. via the catching diode D3.

9.11 C.R.T. Negative EHT is obtained by voltage doubling the 550V AC winding by D7 & D8 and capacitors C43, 44, 45, and 46 both doubling and filtering. Blanking is obtained as discussed in para 9.8 whilst intensity is adjusted by RV6 connected in a divider R92 RV6 R91 RV5 (focus control) and R90 across the EHT supply. Astigmatism is internally preset by RV4.

9.12 POWER SUPPLIES

+50V. A half wave rectified 51V AC winding followed by a three stage filter C53, C52 and C51 supplies all the +50V requirements.

-46V. The same 51V AC winding is also half wave rectified by D12 and followed by C54, 55, and 56 three stage filter for the -ve 46V supply.

The - 54V tapping on the filters used to supply Q9, 10 & 11 in the time base.

+260V. An 82V winding is doubled by D9 and 10, C49 and 50 and added to the +64V rail from D11. A single stage of filtering by R94 C48 is used for the horizontal amplifier and an additional stage R93 C47 for the vertical amplifier supply.

10. ADJUSTMENTS & MAINTENANCE

10.1 A number of preset controls are contained in this instrument which may require periodical adjustments to maintain it is full calibration.

Before removing the top cover disconnect the instrument from the mains. Remove the two screws holding the handle (and 4 screws in the side if fitted) then withdraw the cover. The bottom cover may be removed by unscrewing the feet.

To aid fault finding the voltages present at various points are shown on the circuit.

10.2 Valve replacement will have only a small affect on the calibration of this instrument; however, V1 needs to be reasonably balanced so that the vertical shift control has equal movement either side of centre. If necessary it can be interchanged with V2 or V3.

10.3 bwd 509A ALIGNMENT PROCEDURE

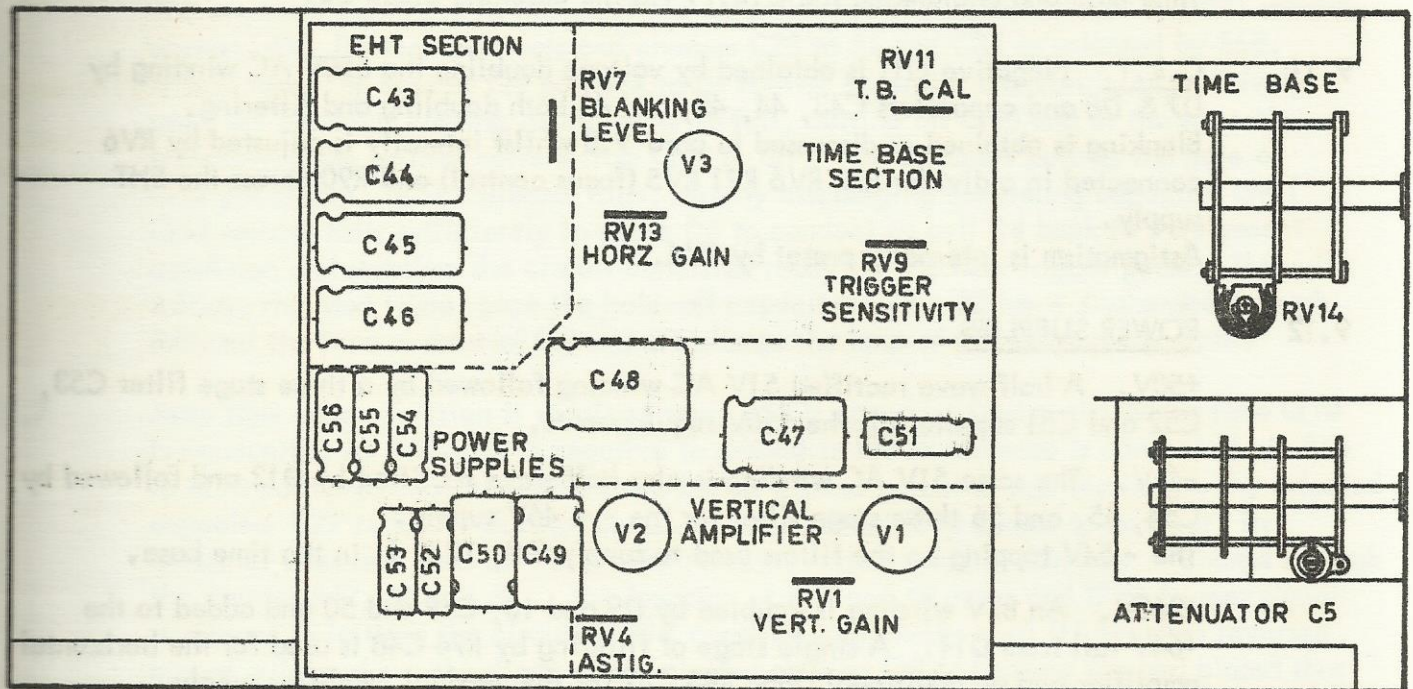
With instrument functioning and trace aligned to graticule, check the following details prior to alignment with Time Base switched to 1mSec.

10.4 ATTEN. to 0.05, centre trace with shift. Turn to 0.1V. If trace moves replace input valve with one free of grid current - type 6ES8.

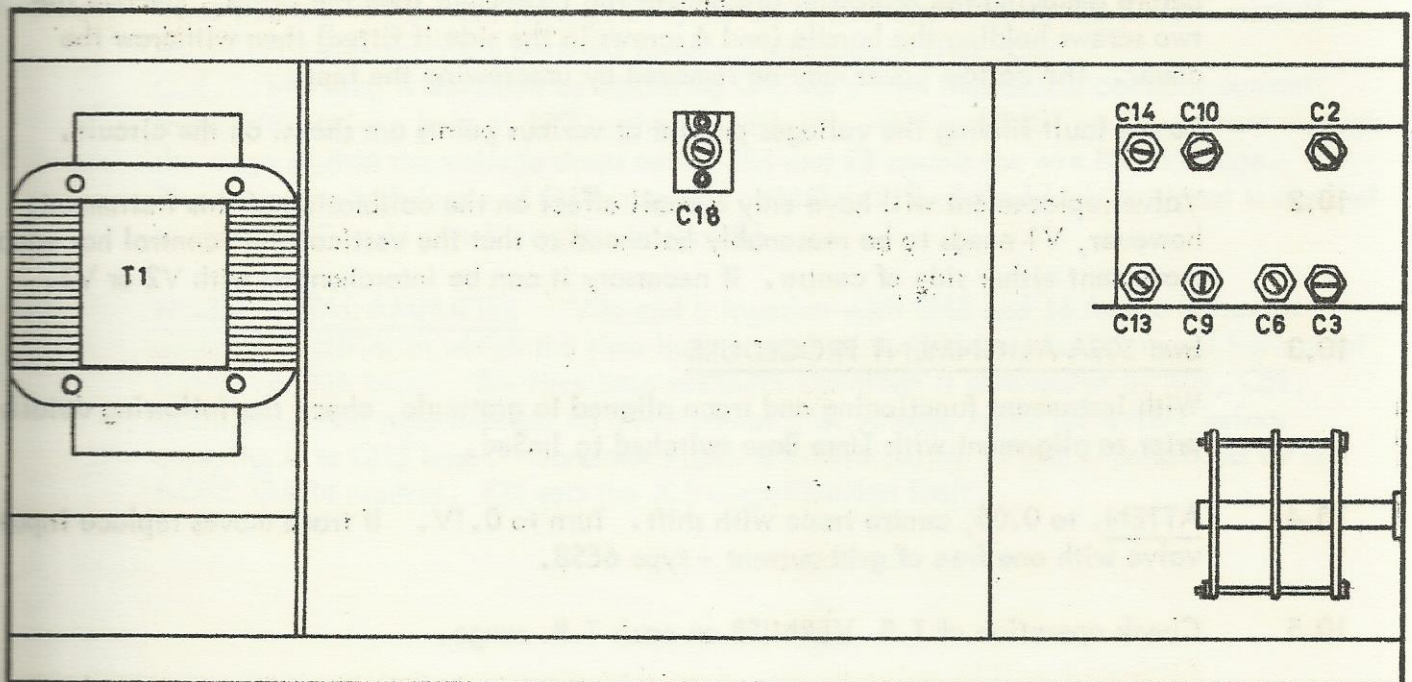
10.5 Check operation of T.B. VERNIER on each T.B. range.

10.6 Turn VERNIER to T.B. OFF, spot should move $\pm 5\text{cm}$ with HORIZONTAL SHIFT.

509A LAYOUT TOP VIEW



BOTTOM VIEW



10.7 General check of controls.

- | | | |
|-----|--------------|---|
| (a) | Intensity: | Complete control over intensity range. |
| (b) | Focus: | Approx. Centre with movement either side. |
| (c) | Horz. Gain: | Trace should expand equally either side of centre |
| (d) | Vert. Shift: | Trace should move completely off screen above and below centre. |

10.8 C.R.T. TRACE ALIGNMENT

If a 1,000Hz sine wave signal is available, feed this into the VERTICAL AMPLIFIER and adjust waveform for 6cms. deflection T.B. to 1mSec. VERNIER at CAL. If a square wave is not available use the CAL waveform, T.B. to 10mSec. VERNIER at CAL.

RV4 on L.H. side rear of P/C Board is adjusted in conjunction with the FOCUS control to obtain the best resolution over the entire screen area when intensity is adjusted to max. brightness but without fly back showing.

10.9 ATTENUATOR & CALIBRATION

Test equipment required 2.5kHz Square Wave Generator.

Set attenuator to 0.05V, feed in 200mV p-p (1% accuracy) Square Wave. Adjust RV1 for 4cm display. Vertical amplifier of Oscilloscope is now correctly calibrated.

The following chart indicates the adjustments necessary to fully align the attenuator.

<u>Attenuator Setting</u>	<u>Input Voltage</u>	<u>Adjust for Square Wave</u>	<u>Input Capacitance</u>
0.05V	200mV	-	-
0.1	500mV	C13	-
0.2	1V	C14	-
0.5	2V	C6	C2
1	5V	C9	-
2	10V	C10	-
5	20V	C5	C3
10	50V	-	-
20	100V	-	-
50	100V	-	-
100	100V	-	-

Attenuator will be automatically aligned at attenuator positions where no capacitor is indicated.

10.10 VERTICAL AMPLIFIER

Test equipment required. 250kHz Square Wave Generator, less than 10 nano Sec. rise time.

Attenuator to 0.05V, Input selector to AC, signal input 200mV p-p 250kHz T.B. range 1 μ Sec, VERNIER to CAL.

Adjust C18 (underside of P/C board) for optimum square wave with minimum over or undershoot.

Check bandwidth, adjust deflection for 4cm at 50kHz, display should not drop to less than 2.8cm at 5MHz.

10.11 HORIZONTAL AMPLIFIER

Test equipment 1Hz to 1MHz Sine Wave Generator, (bwd Model 140A)

Feed in 50kHz sine wave to Vertical Amplifier, Time Base on AUTO at 100 μ Sec/cm, HORZ. GAIN to x 1, Vernier anticlock with time base operating - not at T.B. OFF, Adjust RV13 (located alongside V3 - do not touch RV7 behind it as this component is at -1400V to ground) to set trace length to 10.2cm.

Now disconnect oscillator from Vertical input and reconnect to HORZ. INPUT, Adjust display for 6cm deflection at 1kHz, increase frequency and note frequency when trace drops to 4.2cm length - it should be above 250kHz.

Sensitivity: Feed in 1kHz square wave 5V p-p amplitude, trace should be 10cm long at max gain and 2cm long at X1.

10.12 TRIGGER SENSITIVITY

Feed in 50kHz sine wave, time base to 10 μ Sec/cm, Trigger to AUTO and INT +ve. Reduce amplitude of input signal until trace ceases to lock. Adjust RV9 (centre front of board) for maximum sensitivity of trigger - approx 2 - 3mm display amplitude. Increase display to 1cm deflection increase frequency of input up to 10MHz note trace remains locked both + or -ve selection.

To check low frequency trigger use a bwd 140A oscillator, 1cm deflection will trigger on AUTO to 3Hz. Increase display to 4cm display will now lock to 1Hz.

Trigger Level, Input frequency 1kHz displayed signal 6cm, turn Trigger Level control away from AUTO, trigger point can now be selected over the range ± 3 cm either + or -

10.13 EXTERNAL AUTO Repeat as in para 10.12 but with a 1V p-p input to amplifier linked across to EXT trigger socket. Switch to EXT and check trigger lock over specified range. Increase input to 10V p-p check operation of Trigger Level Control.

10.14 TIME BASE Test equipment required crystal controlled generator with 1 μ Sec to 1 Sec output in decade steps. Set Time Base Range to 1mSec, Vernier to CAL. Horizontal gain to X1. Feed in 1mSec pulse to amplifier and adjust RV11 (T.B. CAL front right of P.C. board) to display 1 pulse per cm. Check the following steps with the frequency indicated and if necessary adjust RV11 for a compromise setting to obtain the minimum error at each step.

<u>T.B. Range</u>	<u>Input Frequency</u>
10 μ Sec	100kHz
100 μ Sec	10kHz
1mSec	1kHz
10mSec	100Hz
100mSec	10Hz (10% accuracy specified)

Now turn to 1 μ Sec and feed in 1 μ Sec pulses. Adjust RV14 located on Time Base switch for one pulse per cm.

Check trace expands approx. X5 and trace remains linear over sweep length (other than first two cm of highest speed (200 nSec/cm.)

- 10.15 C.R.T. BLANKING Care is needed in adjustment of RV7 as this is operating at a potential of -1400V. It is factory adjusted for max. range of the intensity control and should not need adjusting unless the C.R.T. is changed. The adjustment is not critical and should be set with T.B. to 1mSec, AUTO and no input. INTENSITY control anticlock adjust until trace just disappears.

11. REPLACEMENT PARTS

- 11.1 Spares are normally available direct from the manufacturer, B.W.D. ELECTRONICS PTY.LTD. When ordering, it is necessary to indicate the serial number of the instrument. If exact replacements are not to hand, locally available alternatives may be used, provided they possess a specification not less than, or physical size not greater than, the original component.

- 11.2 As the policy of B.W.D. ELECTRONICS PTY.LTD. is one of continuing research and development, the company reserves the right to supply the latest equipment and make amendments to circuits and parts without notice.

12. WARRANTY

- 12.1 The equipment is guaranteed for a period of twelve (12) months from the date of purchase against faulty materials and workmanship, with the exception of cathode ray tubes, which are covered by their manufacturer's own warranty.
- 12.2 Please refer to Guarantee Card No. 4896, which accompanied instrument, for full details of conditions of warranty.

13. ACCESSORIES FOR MODEL bwd-509A OSCILLOSCOPES

High Impedance Probe	Type P22/4mm	(10M Ω and 12pF)
Screened Leads with prods	Type P30	(complete with prod and crocodile clip)
Demodulator Probe	Type P35/4mm	(10kHz to 100MHz 30V max.)

Other associated instruments in the BWD range suitable for use with bwd 509A Oscilloscope are:

bwd 112 Sine and Square Wave Generator 20Hz to 1MHz

bwd 140A Sine and Square Wave Generator 1Hz to 1MHz

bwd 602 Combination instrument, Sine and Square Wave Generators
0.5Hz to 500kHz and multiple stabilised power supplies.

For full details regarding the complete range of BWD Instruments contact the factory address at the front of this Handbook or your nearest distributor.

B.W.D. ELECTRONICS PTY.LTD.

REPLACEABLE PARTS

1. This section contains information for ordering replacement parts, it provides the following details:-
 - (a) Description of part (see list of abbreviations).
 - (b) Typical manufacturer or supplier of the part (see list of abbreviations).
 - (c) Manufacturer's Part Number; and
 - (d) Defence Stock Number, where applicable.
2. Ordering - Please quote Model Type No. e.g. bwd 511, Serial No., Circuit reference No. and component details as listed in parts list.

COMPONENT DESIGNATORS

A	Assembly	H	Heater	RV	Resistor Variable
B	Lamp	J	Jack (socket)	S	Switch
C	Capacitor	L	Inductor	T	Transformer
D	Diode	M	Meter	TH	Thermistor
DL	Delay Line	P	Plug	V	Valve
E	Misc. Elect. Part	Q	Transistor	VDR	Voltage Dependent Resistor
F	Fuse	R	Resistor		

ABBREVIATIONS

Amp	Ampere	L	Inductor
C	Capacitor	lin	Linear
cc	Cracked Carbon	Log	Logarithmic Taper
c	Carbon	m	Milli = 10^{-3}
cd	Deposited Carbon	MHz	Mega Hertz = 10^6 Hz
comp	Composition	MF	Metal Film
CDS	Ceramic Disc Capacitor	ma	Milli Ampere
cer	ceramic	MΩ	Meg Ohm = 10^6 Ω
Com	Common	mfr	Manufacturer
DPST	Double Pole Single Throw	MO	Metal Oxide
DPDT	Double Pole Double Throw	MHT	Polyester/Paper Capacitor
elec	Electrolytic	MPC	Metalised Polyester Capacitor
F	Farad	Ne	Neon
f	Fuse	NPO	Zero temperature co-efficient
FET	Field Effect Transistor	nsr	Not separately replaceable
Ge	Germanium	NC	Normally Closed
H	Henry (ies)	NO	Normally Open
H.S.	High Stability	ns	Nano second
HTC	High Temp Coating	obd	Order by Description
ins	Insulated	OD	Outside Diameter
kHz	Kilo Hertz = 10^3 Hz	p	Peak
KΩ	Kilohm = 10^3 Ω	pf	pico farad = 10^{-12} F

ABBREVIATIONS CONTINUED

PL	Plug	Se	Selenium
PS	Socket	SI	Slide
Preset	Internal Preset	SPDT	Single Pole Double Throw
PYE	Polyester	SPST	Single Pole Single Throw
pot	Potentiometer	Si	Silicon
prec	Precision	Ta	Tantalum
PC	Printed Circuit	tol	Tolerance
PIV	Peak Inverse Voltage	trim	trimmer
PYS	Polystyrene	V	Volt (s)
p-p	Peak to Peak	var	variable
P.Shaft	Plain Shaft	vdw	Volts Direct Current Working
S.Shaft	Slotted Shaft	w	Watt (s)
R	Resistor	ww	Wire Wound
rot	rotary	Z	Zener
R log	Reverse Logarithmic Taper	*	Factory Selected valve average valve may be shown
rms	Root Mean Squared	**	Special Component, no part number assigned

MANUFACTURER ABBREVIATIONS

AC	Allied Capacitors	J	Jabel
AEE	AEE Capacitors	MAS	Master Instrument Co. Pty.Ltd.
AN	Anodeon	MUL	Mullard (Aust.) Pty.Ltd.
AST	Astronic Imports	MOR	Morganite (Aust.) Pty.Ltd.
AWA	Amalgamated Wireless of Aust.	MSP	Manufacturers Special Products (AWA)
ACM	Acme Engineering Pty.Ltd.	McM	McMurdo (Aust.) Pty.Ltd.
AMP	Aircraft Marine Products (Aust)P/L	NU	NU VU Pty.Ltd.
AR	A. & R. Transformers	NAU	A. G. Naunton Pty.Ltd.
AUS	Australux Fuses	PA	Painton (Aust.) Pty.Ltd.
AWV	Amalgamated Wireless Valve Co.	PAL	Paton Elect. Pty.Ltd.
ACA	Amplifier Co. of Aust.	PI	Piher Resistors (Sonar Electronics)
AL	Alpha	PW	Precision Windings Pty.Ltd.
ARR	Arrow	PH	Philips Electrical Industries Pty.Ltd.
BWD	B.W.D. Electronics Pty.Ltd.	PL	Plessey Pacific
BL	Belling & Lee Pty.Ltd.	PV	Peaston Vic.
BR	Brentware (Vic.) Pty.Ltd.	RP	Radio Parts Pty.Ltd.
CF	Carr Fastener	RC	Radio Corporation (Electronic Inds.)
CAN	Cannon Electrics Pty.Ltd.	RCA	Radio Corporation of America
CIN	Cinch	RHC	R. H. Cunningham
D	Ducon Condensor Pty.Ltd.	S	Sonic Electronics Pty.Ltd.
DAR	Darstan	STC	Standard Telephones & Cables
DIS	Distributors Corporation Pty.Ltd.	SI	Siemens Electrical Industries
ELN	Elna Capacitors (Sonar Elect.P/L)	SIM	Simonson Pty.Ltd.
ETD	Electron Tube Dist.	SE	Selectronic Components
F	Fairchild Australia Pty.Ltd.	TR	Trimax Erricson Transformers
GRA	General Radio Agencies	TI	Texas Instruments Pty.Ltd.
GES	General Electronic Services	TH	Thorn Atlas
GL	Grelco	UC	Union Carbide
HW	Hurtle Webster	W	Wellyn Resistors (Cannon Elec.P/L)
HOL	R. G. Holloway	Y	F.L. Yott Pty.Ltd.
H	Haco Distributors (National)	Z	Zephyr Prod. Pty.Ltd.

B.W.D. ELECTRONICS PTY. LTD. - PARTS LIST

MODEL bwd-509A

CCT Ref	DESCRIPTION-RESISTORS				Mfr. or Supplies	PART NO.
R1	900K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R2	990K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R3	111K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R4	9.1K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R5	1K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R6	750K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R7	500K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R8	333K Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R9	1M Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R10	1M Ω	$\frac{1}{4}$ W	1%	HS	ELEC	TR5
R11	33 Ω	$\frac{1}{2}$ W	5%	CC	PI	
R12	56 Ω	$\frac{1}{2}$ W	5%	CC	PI	
R13	220K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R14	82 Ω	$\frac{1}{2}$ W	10%	C	MOR	
R15	100K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R16	3.3K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R17	3.3K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R18	100K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R19	6.8K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R20	2.7K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R21	18K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R22	47K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R23	47K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R24	82 Ω	$\frac{1}{2}$ W	10%	C	MOR	
R25	10K Ω	1W	5%	CC	PI or PH	
R26	10K Ω	1W	5%	CC	PI or PH	
R27	150 Ω	$\frac{1}{2}$ W	5%	CC	PI	
R28	150 Ω	$\frac{1}{2}$ W	5%	CC	PI	
R29	22K Ω	$\frac{1}{2}$ W	5%	CC	PI	
R30						
R31	2.7K Ω	1W	5%	CC	PI or PH	
R32	470K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R33	100K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R34	330K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R35	220K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R36	10K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R37	10K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R38	100K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R39	18K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R40	150K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R41	820 Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R42	10K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	

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CCT Ref	DESCRIPTION - RESISTORS				Mfr. or Supplies	PART No.
R43	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R44	27K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R45	3.3K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R46	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R47	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R48	2.2K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R49	220K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R50	6.8K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R51	220 Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R52	47K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R53	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R54	15K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R55	120K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R56	47K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R57	1K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R58	22K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R59	56K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R60	22K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R61	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R62	47K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R63	1.5M Ω	$\frac{1}{2}$ W	5%	CC	PI	
R64	10M Ω	$\frac{1}{2}$ W	5%	CC	PI	
R65	3.3M Ω	$\frac{1}{2}$ W	5%	CC	PI	
R66	33K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R67	150 Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R68	2.7K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R69	330K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R70	12K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R71	18K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R72	470K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R73	47K Ω	1W	5%	CC	PI or PH	
R74	22K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R75	680 Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R76	68K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R77	22K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R78	47K Ω	1W	5%	CC	PI or PH	
R79	22K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R80	100K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R81	56K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R82	10K	$\frac{1}{2}$ W	5%	CC	PI or PH	
R83	47K Ω	$\frac{1}{2}$ W	5%	CC	PI or PH	
R84	1M Ω	1W	5%	CC	PI or PH	
R85	1M Ω	1W	5%	CC	PI or PH	

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CCT Ref	DESCRIPTION — RESISTORS				Mfr. or Supplier	PART No.
R86	1MΩ	½W	5%	CC	PI or PH	
R87	1MΩ	½W	5%	CC	PI or PH	
R88	1MΩ	½W	5%	CC	PI or PH	
R89	1MΩ	½W	5%	CC	PI or PH	
R90	1.8MΩ	1W	10%	CC	PI or PH	
R91	470KΩ	½W	5%	CC	PI or PH	
R92	100KΩ	½W	5%	CC	PI or PH	
R93	1KΩ	1W	5%	CC	PI or PH	
R94	1KΩ	1W	5%	CC	PI or PH	
R95	470Ω	½W	5%	CC	PI or PH	
R96	680Ω	½W	5%	CC	PI or PH	
R97	470Ω	½W	5%	CC	PI or PH	
R98	560Ω	½W	5%	CC	PI or PH	
R99	1MΩ	½W	5%	CC	PI or PH	
R100	3.3K	½W	5%	CC	PI or PH	
R101	6.8KΩ	½W	5%	CC	PI or PH	
R102	220KΩ	½W	5%	CC	PI or PH	
CAPACITORS						
C1	.1μF	400V	10%	PYE	PH	C296AC/A100K
C2	2.20 pF	trim			D	CTS2-20
C3	2.20 pF	trim			D	CTS2-20
C4	5.6pF	500V	±1pF	NPO	D or AC	CDS
C5	.7-3pF	trim			PH	C004AA/3E
C6	1-12pF	trim			PH	C004CA/12E
C7	470pF	500V	MICA	5%	D	MSA
C8	.0047pF	400V	10%	PYE	PH	C296AC/A4K7
C9	1-12pF	trim			PH	C004CA/12E
C10	2-20pF	trim			D	CTS2-20
C11	8.2pF	500V	±1pF	NPO	D or AC	CDS
C12	27pF	500V	5%	N330	D or AC	CDS
C13	1-12pF	trim			PH	C004CA/12E
C14	1-12pF	trim			PH	C004CA/12E
C15	.01μF	100V	10%	MPC	ELNA	GREENCAP
C16	39pF	500V	5%	N750	D or AC	CDS
C17						
C18	20-220pF	trim			D	CWO
C19	680pF	630V	10%	PYS	AC	TCS610
C20	3.3pF	500V	±½pF	NPO	AC or D	CDS
C21	.22 uF	100V	10%	MPC	ELNA	GREENCAP
C22	100pF	500V	10%	N750	D or AC	CDS
C23	1μF	200V	10%	MPC	ELNA	GREENCAP

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MODEL bwd-509A

CCT Ref	DESCRIPTION-CAPACITORS				Mfr. or Supplies	PART No.
C24	22pF	500V	5%	N330	D or AC	CDS
C25	10pF	500V	5%	NPO	D or AC	CDS
C26	22pF	500V	5%	N330	D or AC	CDS
C27	10 μ F	64V	ELEC		PH	C426AR/H10
C28	.64 μ F	64V	ELEC		PH	C426AR/H064
C29	.047	160V	10%	PYE	PH	C296AC/47K
C30	.0047	400V	10%	PYE	PH	C296AC/4K7
C31	390pF	630V	10%	PYS	AC	TCS607
C32	220pF	630V	10%	PYS	AC	TCS604
C33	22pF	500V	5%	N330	AC or D	CDS
C34	.01 μ F	100V	10%	MPC	ELNA	GREENCAP
C35	.1 μ F	100V	2%	SELECTED MPC	ELNA	GREENCAP
C36	.01 μ F	100V	2%	SELECTED MPC	ELNA	GREENCAP
C37	.001 μ F	630V	2%	PYS	AC	TCS412
C38	95pF	500V	5%	N750	AC or D	CDS
C39	100pF	500V	10%	N750	AC or D	CDS
C40	.1 μ F	100V	10%	MPC	ELNA	GREENCAP
C41	10 μ F	25V	ELEC		PH	C426AR/F10
C42	.01 μ F	3KV	CER DISC		D	CDH
C43	8 μ F	450V	ELEC		ELNA	
C44	8 μ F	450V	ELEC		ELNA	
C45	8 μ F	450V	ELEC		ELNA	
C46	8 μ F	450V	ELEC		ELNA	
C47	16 μ F	350V	ELEC		ELNA	
C48	32 μ F	300V	ELEC		D	ET5B
C49	50 μ F	150V	ELEC		PH	C436AR/K50
C50	50 μ F	150V	ELEC		PH	C436AR/K50
C51	64 μ F	64V	ELEC		PH	C437AR/H64
C52	64 μ F	64V	ELEC		PH	C437AR/H64
C53	64 μ F	64V	ELEC		PH	C437AR/H64
C54	64 μ F	64V	ELEC		PH	C437AR/H64
C55	64 μ F	64V	ELEC		PH	C437AR/H64
C56	64 μ F	64V	ELEC		PH	C437AR/H64
C57	220pF	2000V	$\pm 20\%$		D	CDH
C58	39pF	500V	5%	N750	AC or D	CDS
C59	10pF	500V	$\pm 1pF$	NPO	AC or D	CDS
C60	180pF	630V	5%	PYS	AC	TCS603
C61	15pF	500V	5%	N330	AC or D	CDS
C62	3.3pF	500V	$\pm 1pF$	NPO	AC or D	CDS
C63	5.6pF	500V	$\pm 1pF$	NPO	AC or D	CDS

B.W.D. ELECTRONICS PTY.LTD. - PARTS LIST

MODEL bwd-509A

CCT Ref	DESCRIPTION - TRANSISTORS				Mfr. or Supplies	PART No.
Q1,2,3,&4	BF194	SI			PH	BF194
Q5,6,7	BC147	SI			PH	BC147
Q9,10,11	BC147	SI			PH	BC147
Q13,14,15	BC147	SI			PH	BC147
Q16,17	BC147	SI			PH	BC147
Q8 & 11	BC157	SI			PH	BC147
Q12	2N3819	SELECTED FET			TI	2N3819
	<u>DIODES</u>					
D1,2,3,	OA91	Ge			PH	OA91
D5,6	OA91	Ge			PH	OA91
D4	IN914	SI			AWV	IN914
D7	K8/25	RECTIFIER	Se		STC	K8/25
D8	K8/25	RECTIFIER	Se		STC	K8/25
D9	EM404 or SD55/4		SI		STC	EM404
D10	EM404 or SD55/4		SI		STC	EM404
D11	EM402 or SD55/4		SI		STC	EM402
D12	EM402 or SD55/4		SI		STC	EM402
	<u>VALVES</u>					
V1	6ES8				PH	6ES8
V2	6ES8				PH	6ES8
V3	6ES8				PH	6ES8
	<u>CRT</u>					
V4	5UPI-F or D13-480GM					
	<u>FUSE</u>					
F1	0.5A QUICK BLOW				Y	3A6
	<u>POTENTIOMETER</u>					
RV1	100K "A"	Preset	C		PH	EO97AC/100K
RV2	500K "A"	Preset	C		ELNA	
RV3						
RV4	100K "A"	Preset	C		PH	EO97AC/100K
RV5	1MΩ "A"		C		ELNA	
RV6	250K "A" with DPST ROTARY SW				ELNA	
RV7	4.7K "A"	Preset	C		PH	EO97AC/4K7
RV8	250K "A" with DPST ROTARY SW				ELNA	
RV9	22KΩ "A"	Preset	C		PH	
RV10	250K "A" with DPST ROTARY SW				ELNA	
RV11	100K "A"	Preset	C		PH	EO97AC/100K
RV12A	100K "A" REAR SECTION		C		D)	PDU
RV12B	10K "F" FRONT SECTION		C		D)	PDU
RV13	10K "A"	Preset	C		PH	EO97AC/10K

B.W.D. ELECTRONICS PTY.LTD. - PARTS LIST

MODEL bwd-509A

CCT Ref	DESCRIPTION — POTENTIOMETER			Mfr. or Supplier	PART No.
RV 14	220K Ω "A"	Preset	C	PH	EO97AC/220K
VDR	E299 DD/234			PH	E299 DD/234

MODIFICATION

ISSUE 1 7-69

ISSUE 2 8-69

ISSUE 3 10-69

PRODUCTION

SWITCHES

S1 AC-DC INPUT

S2A-D INPUT ATTENUATOR

S3 INT-EXT. TRIGGER

S4A&B + OR - TRIGGER

S5A&B AUTO OR LEVEL SELECT [REAR RV8]

S6A-D TIME BASE RANGE

S7A&B INT-EXT. TIME BASE [REAR RV10]

S8 A&B AC POWER ON-OFF [REAR RV6]

CONTROLS

RV1 VERT GAIN CALIBRATE

RV2 VERT POSITION

RV4 ASTIGMATISM

RV5 FOCUS

RV6 INTENSITY

RV7 BLANKING PRESET

RV8 TRIGGER LEVEL

RV9 TRIGGER PRESET

RV10 TIME BASE VERNIER

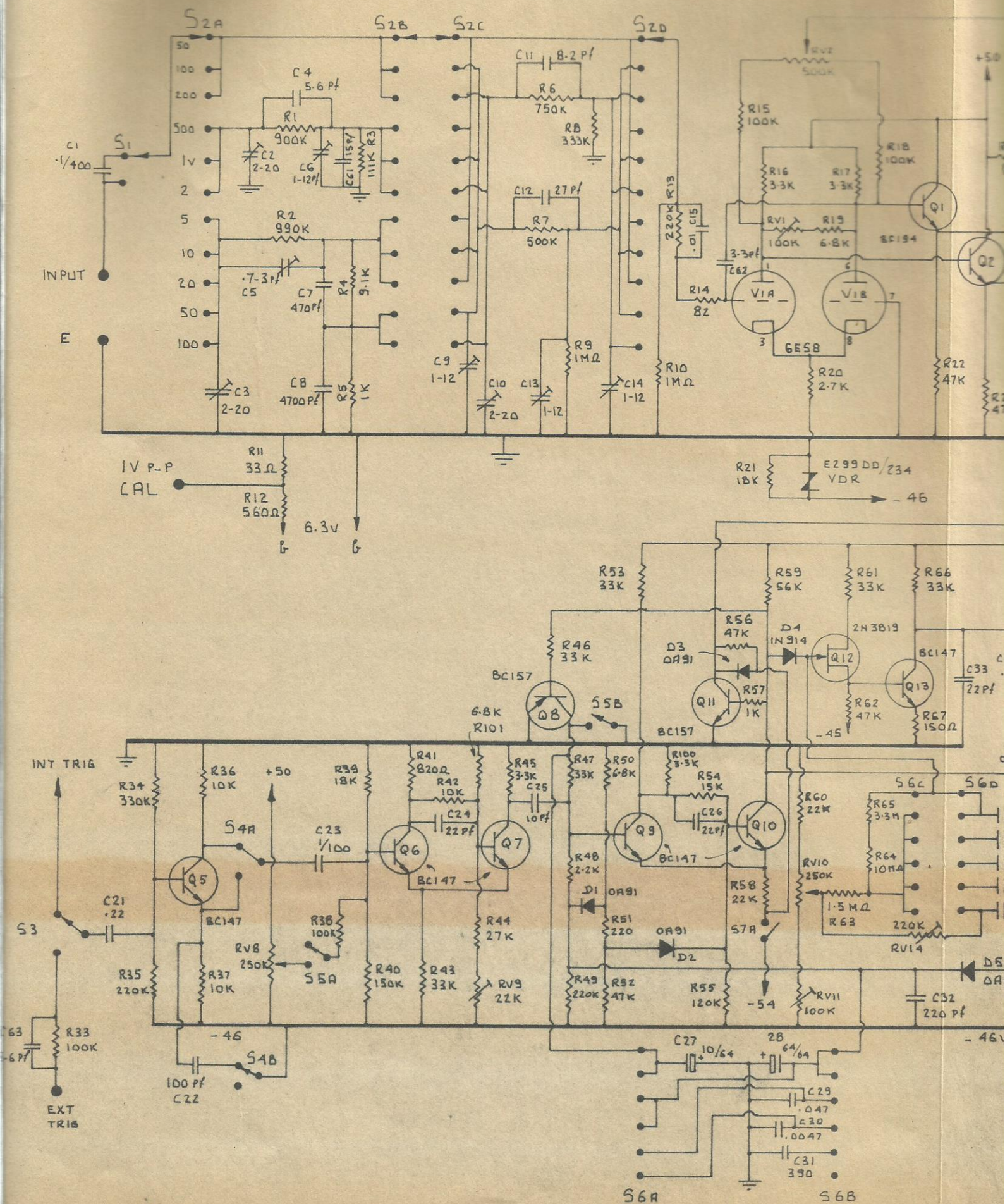
RV11 TIME BASE CALIBRATE

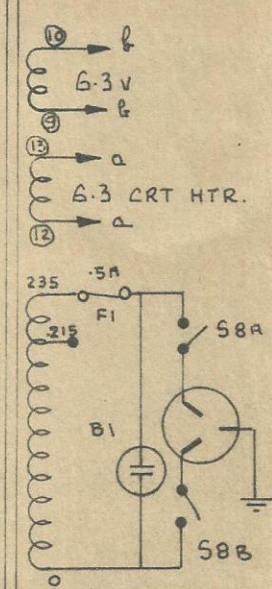
RV12A HORIZ. SHIFT

RV12B HORIZ. MAGNIFICATION

RV13 HORIZ. " X1 PRESET

RV14 14 SEC TB RANGE CAL.





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